# Fully Integrated Linear Single Cell Li-Ion Battery Charger with Power Path

### **General Description**

The RT9503A is a fully integrated low cost single-cell Lilon battery charger IC ideal for portable applications. The RT9503A is capable of being powered up via the AC adapter and USB (Universal Serial Bus) port inputs. The RT9503A can automatically detect and select the AC adapter and the USB port as the power source for the charger. The RT9503A enters sleep mode when both supplies are removed.

The RT9503A optimizes the charging task by using a control algorithm including preconditioning mode, fast charge mode, and constant voltage mode. The charging task is terminated as the charge current drops below the preset threshold. The USB charge current can be selected from preset ratings of 100mA and 500mA, while the AC adapter charge current can be programmed up to 1A with an external resistor. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures.

The RT9503A features 18V and 7V maximum rating voltages for AC adapter and USB port inputs respectively. Other features include external programmable safety timer, under voltage protection, over voltage protection for AC adapter supply, battery temperature monitoring, power supply status indicators and charge status indicator.

### **Ordering Information**

RT9503A 🗖 📮

Package Type
QW : WQFN-16L 3x3 (W-Type)
Lead Plating System
P : Pb Free
G : Green (Halogen Free and Pb Free)

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

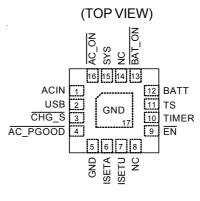
### Features

- Automatic Input Supplies Selection
- 18V Maximum Rating for AC Adapter
- Integrated Selectable 100mA and 500mA USB Charge Current
- Internal Integrated Power MOSFETs
- AC Adapter Power Good Status Indicator
- Charge Status Indicator
- External Capacitor Programmable Safety Timer
- Under Voltage Protection
- Over Voltage Protection
- Automatic Recharge Feature
- Battery Temperature Monitoring
- Small 16-Lead WQFN Package
- Thermal Feedback Optimizing Charge Rate
- Power Path Controller
- RoHS Compliant and Halogen Free

### Applications

- Digital Cameras
- Cellular Phones
- PDAs , Smart Phones and MP3 Players
- Portable Instruments

# **Pin Configurations**



WQFN-16L 3x3

# RT9503A

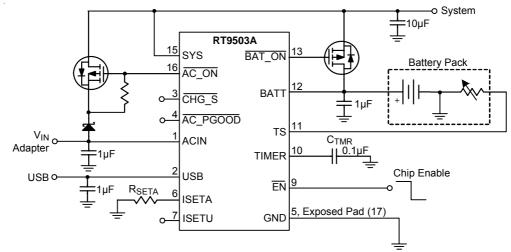


# **Marking Information**



HK= : Product Code YMDNN : Date Code

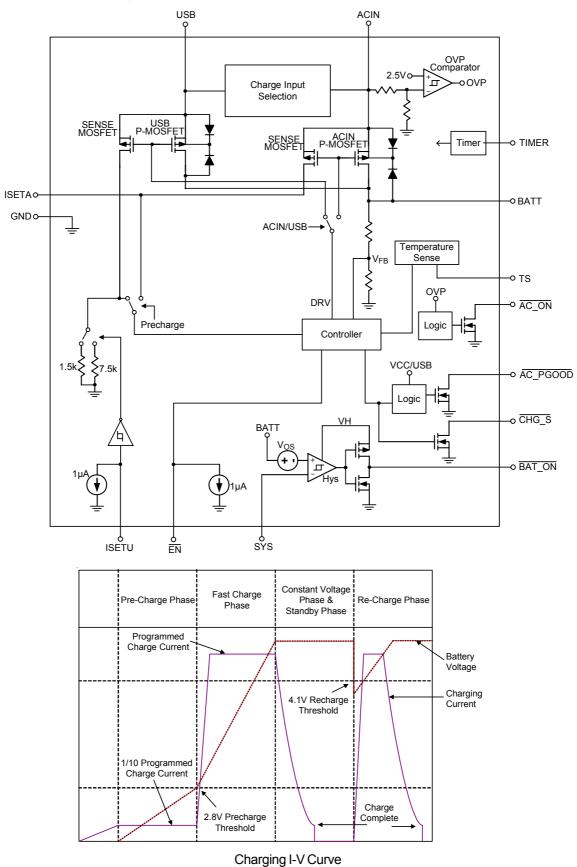
# **Typical Application Circuit**



# **Functional Pin Description**

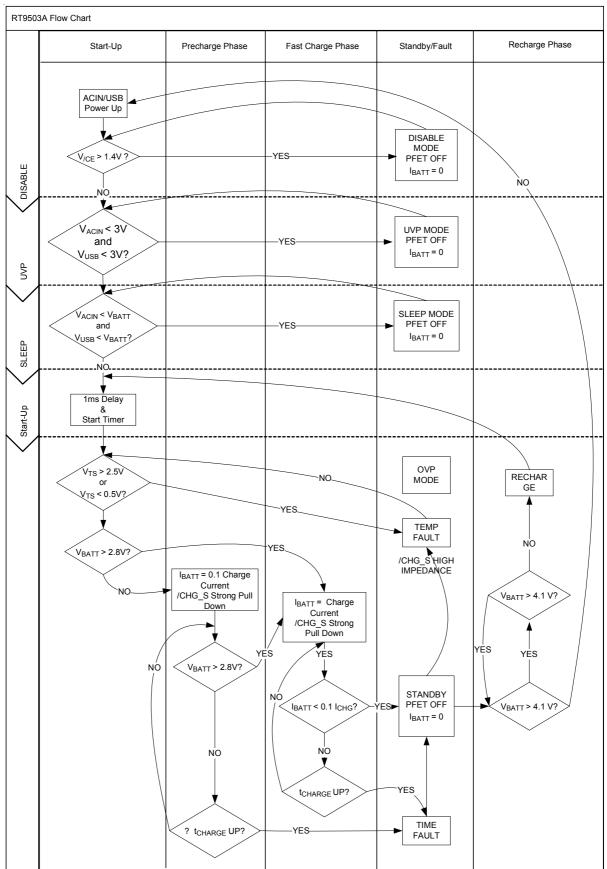
Pin No.	Pin Name	Pin Function	
1	ACIN	Wall Adaptor Charge Supply Input.	
2	USB	USB Charge Supply Input.	
3	CHG_S	Charge Status Indicator Output (open drain).	
4	AC_PGOOD	Wall Adaptor Power Good Indicator Output (open drain).	
5, 17 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.	
6	ISETA	Wall Adaptor Supply Charge Current Set Point.	
7	ISETU	USB Supply Charge Current Set Input (active low).	
8, 14	NC	No Internal Connection.	
9	EN	Charge Enable (Active Low).	
10	TIMER	Safe Charge Timer Setting.	
11	TS	Temperature Sense Input.	
12	BATT	Battery Charge Current Output.	
13	BAT_ON	Power Path Controller Output. Pull low to turn on the external P-MOSFET.	
15	SYS	System Voltage Detecting Input.	
16	AC_ON	P-MOSFET Switch Control Output (open drain).	

# **Function Block Diagram**









# Absolute Maximum Ratings (Note 1)

_	
Supply Input Voltage, ACIN	–0.3V to 18V
Supply Input Voltage, USB	–0.3V to 7V
• CHG_S, AC_PGOOD, AC_ON	–0.3V to 18V
Other Pins	0.3V to 5.5V
• Power Dissipation, $P_D @ T_A = 25^{\circ}C$	
WQFN-16L 3x3	1.471W
Package Thermal Resistance (Note 2)	
WQFN-16L 3x3, θ <sub>JA</sub>	68°C/W
WQFN-16L 3x3, θ <sub>JC</sub>	7.5°C/W
• Lead Temperature (Soldering, 10 sec.)	260°C
• Junction Temperature	150°C
Storage Temperature Range	–65°C to 150°C
ESD Susceptibility (Note 3)	
HBM (Human Body Mode)	2kV
MM (Machine Mode)	200V

### Recommended Operating Conditions (Note 4)

Supply Input Voltage Range, ACIN	- 4.5V to 12V
Supply Input Voltage Range, USB	- 4.1V to 6V
Junction Temperature Range	- –20°C to 125°C
Ambient Temperature Range	20°C to 85°C

### **Electrical Characteristics**

(ACIN = USB = 5V,  $T_A$  = 25°C, unless otherwise specification)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit	
Supply Input							
ACIN UVP Rising Threshold Voltage	V <sub>UV_ACIN</sub>			3		V	
USB UVP Rising Threshold Voltage	V <sub>UV_USB</sub>	V <sub>BATT</sub> = 3V		3		V	
ACIN/USB UVP Hysteresis	V <sub>UV_HYS</sub>	V <sub>BATT</sub> = 3V		100		mV	
ACIN/USB Standby Current	I <sub>STBY</sub>	V <sub>BATT</sub> = 4.5V		300	500	μA	
BATT Sleep Leakage Current	I <sub>SLEEP</sub>	$V_{ACIN} = 4V, V_{USB} = 4V,$ $V_{BATT} = 4.5V$		5	15	μA	
Voltage Regulation							
BATT Regulation Voltage	V <sub>REG</sub>	I <sub>BATT</sub> = 60mA	4.138	4.2	4.262	V	
ACIN MOSFET Dropout		V <sub>BATT</sub> = 4V, I <sub>CHG_AC</sub> = 1A	400	500	620	mV	
USB MOSFET Dropout		$V_{BATT}$ = 4V, $I_{CHG}_{USB}$ = High	500	650	800	mV	
Current Regulation							
ISETA Set Voltage (Fast Charge Phase)	VISETA_FCHG	V <sub>BATT</sub> = 3.5V	2.43	2.48	2.53	V	
Full Charge Setting Range	I <sub>CHG_AC</sub>		100		1000	mA	
AC Charge Current Accuracy	I <sub>CHG_AC</sub>	$V_{BATT}$ = 3.8V, $R_{ISET}$ = 1.5k $\Omega$		500		mA	

To be continued

# RT9503A



Parameter		Symbol	Test Conditions	Min	Тур	Мах	Unit
Precharge							
BATT Pre-charge Threshold		V <sub>PRECH</sub>		2.7	2.8	2.9	V
BATT Pre-charge 1 Hysteresis	hreshold	$\Delta V_{PRECH}$		60	100	140	mV
Pre-Charge Curren	t	I <sub>PCHG</sub>	V <sub>BATT</sub> = 2V	8	10	12	%
Recharge Thresh	old						
BATT Re-charge F Hysteresis	alling Threshold	$\Delta V_{\text{RECH_L}}$		50	95	140	mV
Charge Terminati	on Detection						
Termination Currer	nt Ratio (Note5)	ITERM	V <sub>BATT</sub> = 4.2V		10		%
Logic Input/Outpu	ıt						
CHG_S Pull Down	Voltage	V <sub>CHG_S</sub>	I <sub>CHG_S</sub> = 5mA		213		mV
AC_PGOOD Pull D	own Voltage	VPGOOD	I <sub>PGOOD</sub> = 5mA		213		mV
PGOOD Pull Down	Voltage	Vpgood	IPGOOD = 5mA		65		mV
EN Threshold	Logic-High	VIH		1.5			V
Voltage	Logic-Low	VIL				0.4	V
EN Pin Input Curre	nt	IEN				1.5	μA
	High Voltage	VISETU_HIGH		1.5			V
ISETU Threshold Low Voltage		VISETU_LOW				0.4	V
ISETU Pin Input Current		IISETU				1.5	μA
USB Charge Curr	ent & Timing						
Soft-Start Time		t <sub>SS</sub>	VISETA from 0V to 2.5V		100		μS
USB Charge Current		ICHG_USB	V <sub>ACIN</sub> = 2.5V, V <sub>USB</sub> = 5V, V <sub>BATT</sub> =3.5V, V <sub>ISETU</sub> = 5V	400	450	500	mA
USB Charge Current		I <sub>CHG_USB</sub>	V <sub>ACIN</sub> = 2.5V, V <sub>USB</sub> = 5V, V <sub>BATT</sub> = 3.5V, V <sub>ISETU</sub> = 0V	60	80	100	mA
Timer		•					
TIMER Pin Source	Current	ITIME	V <sub>TIMER</sub> = 2V		1		μA
Pre-charge Fault Ti	me	t <sub>PCHG_F</sub>	$C_{TIMER}$ = 0.1 $\mu$ F, f <sub>CLK</sub> = 7Hz	1720	2460	3200	S
Charge Fault Time		t <sub>FCHG_F</sub>	$C_{\text{TIMER}}$ = 0.1µF, f <sub>CLK</sub> = 7Hz	13790	19700	25610	S
Battery Temperate	ure Sense						
TS Pin Source Cur	rent	I <sub>TS</sub>	V <sub>TS</sub> = 1.5V		102		μA
TS Pin High	Voltage	V <sub>TS_HIGH</sub>		0.485	0.5	0.515	V
Threshold Low Voltage		V <sub>TS_LOW</sub>		2.45	2.5	2.55	V
Protection							
Thermal Regulation					125		°C
OVP SET Voltage			Internal Default		6.5		V
Power Path Contr	oller						
BAT_ON Pull Low			As SYS Falling, V <sub>BATT</sub> = 4V, SYS-BAT	-150		-20	mV

To be continued

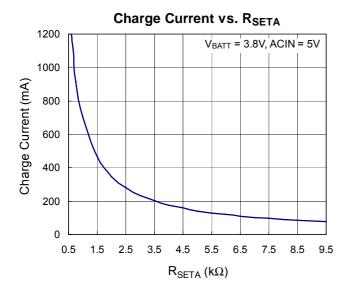
# RT9503A

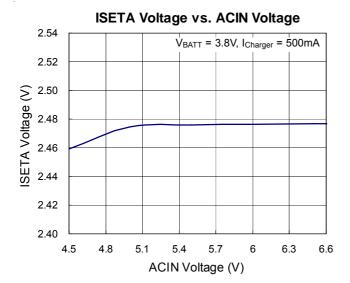
Parameter	Symbol	Test Conditions	Min	Тур	Мах	Uni
BAT_ON Pull High		As SYS Raising, V <sub>BATT</sub> = 4V, SYS-BAT	-50		0	mV
BAT_ON Pull Low Switch Resistance		V <sub>BATT</sub> = 4V		10		Ω
BAT_ON Pull High Switch Resistance		V <sub>ACIN</sub> = 5V		30		Ω

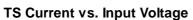
- **Note 1.** Stresses listed as the above "Absolute Maximum Ratings" may cause permanent damage to the device. These are for stress ratings. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may remain possibility to affect device reliability.
- **Note 2.**  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25^{\circ}C$  on a high effective thermal conductivity four-layer test board of JEDEC 51-7 thermal measurement standard. The measurement case position of  $\theta_{JC}$  is on the exposed pad of the package.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.
- Note 5. Guaranteed by design.

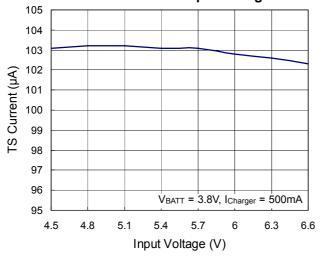


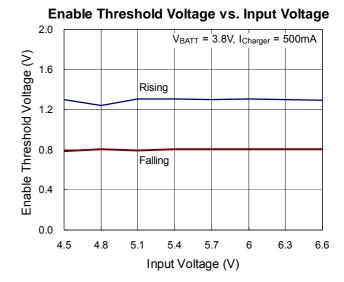
### **Typical Operating Characteristics**



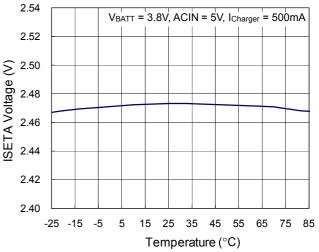




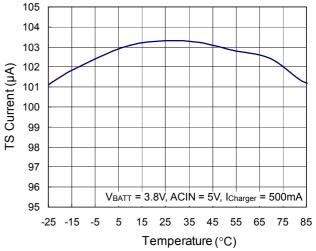




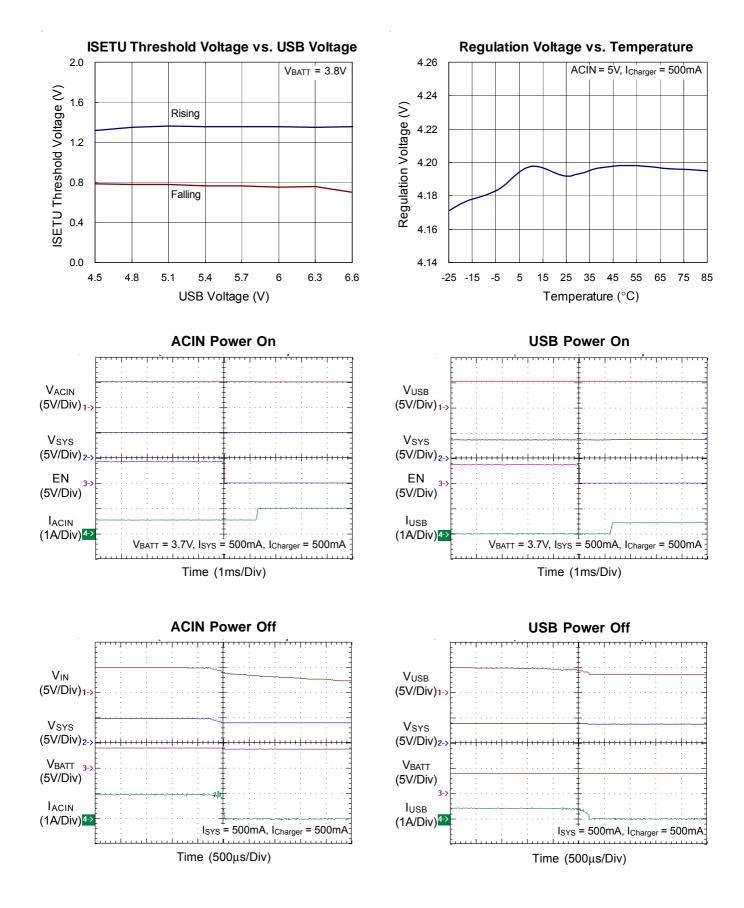
**ISETA Voltage vs. Temperature** 



TS Current vs. Temperature



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### **Application Information**

The RT9503A is a fully integrated low cost single-cell Lilon battery charger for portable applications. The RT9503A can be adopted for two input power source including AC and USB inputs. It will automatically select the input source and operate in different mode as below.

AC Mode : When the AC input voltage (ACIN) is higher than the UVP voltage level (3V), the RT9503A will enter AC Mode. In the AC Mode, ACIN P-MOSFET is turned on and USB P-MOSFET is turned off. When ACIN voltage is below and OVP threshold levels, the switch Q1 will be turned on and Q2 will be turned off. So, the system load is powered directly from the adapter through the transistor Q1, and the battery is charged by the RT9503A. Once the ACIN voltage is higher than the OVP, the RT9503A stops charging, and then Q1 will be turned off and Q2 will be turned on to supply the system by battery.

USB Mode : When AC input voltage (ACIN) is removed and USB input voltage is higher than UVP voltage level (3V), the RT9503A will operate in the USB Mode. In the USB Mode, ACIN P-MOSFET is turned off while USB P-MOSFET and Q2 are turned on. The system load is powered directly from the USB/Battery through the switch Q2. Note that in this mode, the battery will be discharged once the system current is higher than the battery charge current.

Sleep Mode : The RT9503A will enter Sleep Mode when both AC and USB input voltage are removed. This feature provides low leakage current from the battery during the absence of input supply.

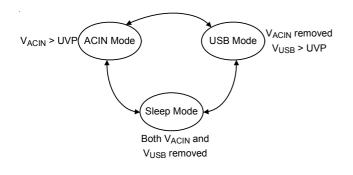


Figure 1. Input Power Source Operation Mode

#### **Power-Path Management**

The RT9503A powers the system and independently charging the battery while the input source is AC. This feature reduces the charge time, allows for proper charge termination, and allows the system to run with an absent or defective battery pack.

#### Case 1 : Input Source is AC

In this case, the system load is powered directly from the AC adapter through the transistor Q1. For RT9503A, Q1 and Q2 act as a switch as long as the RT9503A is ready. Once the AC voltage is ready (<OVP), the battery charge by the RT9503A internal MOSFET and Q1 starts regulating the output voltage supply system (Q2 is turn off). Once the AC voltage is higher than the operation voltage (>OVP), the RT9503A stop charging battery, Q1 turns off and Q2 starts to supply power for system.

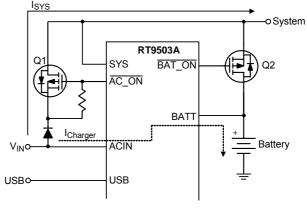
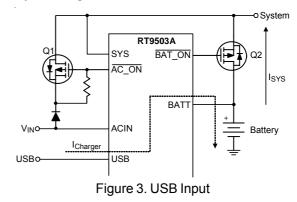


Figure 2. ACIN Input

### Case 2 : Input Source is USB

In this case, the system load is powered directly from the battery through the switch Q2. Note that in this case, the system current over battery charge current will lead to battery discharge.



# RT9503A

#### **ACIN Over-Voltage Protection**

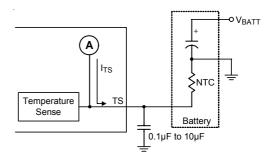
The ACIN input voltage is monitored by an internal OVP comparator. The comparator has an accurate reference of 2.5V from the band-gap reference. The OVP threshold is set by the internal resistor. The protection threshold is set to 6.5V, but ACIN input voltage over 18V still leads the RT9503A to damage. When the input voltage exceeds the threshold, the comparator outputs a logic signal to turn off the power P-MOSFET to prevent the high input voltage from damaging the electronics in the handheld system. When the input over voltage condition is removed (ACIN < 6V), the comparator re-enables the output by running through the soft-start.

#### **Battery Temperature Monitoring**

The RT9503A continuously monitors battery temperature by measuring the voltage between the TS and GND pins.

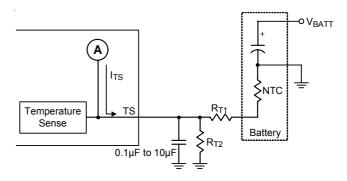
The RT9503A has an internal current source to provide the bias for the most common  $10k\Omega$  negative-temperature coefficient thermal resistor (NTC) (see Figure 4). The RT9503A compares the voltage on the TS pin against the internal  $V_{TS HIGH}$  and  $V_{TS LOW}$  thresholds to determine if charging is allowed. When the temperature outside the  $V_{\text{TS HIGH}}$  and  $V_{\text{TS LOW}}$  thresholds is detected, the device will immediately stop charging. The RT9503A stops charging and keeps monitoring the battery temperature when the temperature-sense voltage is back to the threshold between  $V_{TS HIGH}$  and  $V_{TS LOW}$ , the charger will be resumed. Charge is resumed when the temperature returns to the normal range. However, the user may modify thresholds via the negative-temperature coefficient thermal resistor or by adding two external resistors. (see Figure 5.)

The capacitor should be placed close to TS (Pin 11) and connected to the ground plane. The capacitance value (0.1 $\mu$ F to 10 $\mu$ F) should be selected according to the quality of PCB layout. It is recommended to use 10 $\mu$ F if the layout is poor for noise prevention.



 $\label{eq:VTS} \begin{array}{l} V_{TS} \mbox{ = } I_{TS} \times R_{NTC} \\ Turn \mbox{ off when } V_{TS} \geq 2.5 V \mbox{ or } V_{TS} \leq 0.5 V \end{array}$ 

Figure 4. Temperature Sensing Configuration



$$\label{eq:VTS} \begin{split} &\mathsf{V}_{TS} = \mathsf{I}_{TS} \, \frac{\mathsf{R}_{T2} \times (\mathsf{R}_{T1} + \mathsf{R}_{NTC})}{\mathsf{R}_{T1} + \mathsf{R}_{T2} + \mathsf{R}_{NTC}} \\ &\mathsf{Turn off when } \mathsf{V}_{TS} \geq 2.5 \mathsf{V} \text{ or } \mathsf{V}_{TS} \leq 0.5 \mathsf{V} \end{split}$$

Figure 5. Temperature Sensing Circuit

#### Fast-Charge Current Setting

#### Case 1: ACIN Mode

The ISETA pin is used to determine the ACIN charge rate from 100mA to 1A. The charge current can be calculated as following equation.

$$I_{CHG_AC} = K_{SET} \frac{V_{SET}}{R_{SETA}}$$

The parameter  $K_{SET}$  = 300 ;  $V_{SET}$  = 2.5V.  $R_{SETA}$  is the resistor connected between ISETA and GND.

# RT9503A

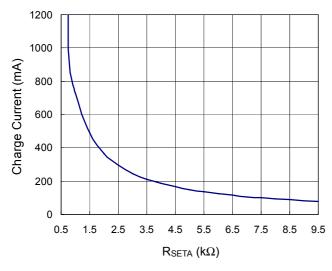


Figure 6. AC Mode Charge Current Setting

### Case 2 : USB Mode

When charging from a USB port, the ISETU pin can be used to determine the charge current of 100mA or 500mA.

A low-level signal of ISETU pin sets the charge current at 100mA and a high level signal sets the charge current at 500mA.

#### **Pre-Charge Current Setting**

During a charge cycle, if the battery voltage is below the  $V_{PRECH}$  threshold, the RT9503A applies a pre-charge mode to the battery. This feature revives deeply discharged cells and protects battery life. The RT9503A internally determines the pre-charge rate as 10% of the fast-charge current.

### **Battery Voltage Regulation**

The RT9503A monitors the battery voltage through the BATT pin. Once the battery voltage level closes in on the V<sub>REG</sub> threshold, the RT9503A voltage enters constant phase and the charging current begins to taper down. When battery voltage is over the V<sub>REG</sub> threshold, the RT9503A will stop charging but continue to monitor the battery voltage. However, when the battery voltage decreases 100mV below V<sub>REG</sub>, it will be recharged to preserve the battery voltage.

### **Charge Status Outputs**

The open-drain CHG\_S and AC\_PGOOD outputs indicate various charger operations as shown in the following table. These status pins can be used to drive LEDs or

communicate to the host processor. Note that ON indicates the open-drain transistor is turned on and LED bright.

Charge State		CHG_S	AC_PGOOD
ACIN	Charge	ON	ON
	Charge Done	OFF	ON
USB	Charge	ON	OFF
036	Charge Done	OFF	OFF

#### **Temperature Regulation and Thermal Protection**

In order to maximize the charge rate, the RT9503A features a junction temperature regulation loop. If the power dissipation of the IC results in a junction temperature greater than the thermal regulation threshold ( $125^{\circ}C$ ), the RT9503A throttles back on the charge current in order to maintain a junction temperature around the thermal regulation threshold ( $125^{\circ}C$ ). The RT9503A monitors the junction temperature, T<sub>J</sub>, of the die and disconnects the battery from the input if T<sub>J</sub> exceeds  $125^{\circ}C$ . This operation continues until junction temperature falls below thermal regulation threshold ( $125^{\circ}C$ ) by the hysteresis level. This feature prevents the maximum power dissipation from exceeding typical design conditions.

### **External Timer**

As a safety mechanism, the RT9503A provides a userprogrammable timer that monitors the pre-charge and fast charge time. This timer (charge safety timer) is started at the beginning of the pre-charge and fast charge period. The safety charge timeout value is set by the value of an external capacitor connected to the TIMER pin ( $C_{TMR}$ ), if the TIMER pin is short to GND, the charge safety timer is disabled.

As  $C_{TMR}$  = 0.1µF,  $t_{PCHG_F}$  is ~2460 secs and  $t_{FCHG_F}$  is 8 x  $t_{PCHG_F}$ .  $t_{PCHG_F}$  =  $C_{TMR}$  x 2460/0.1µ

When timer faults, re-plug-in power or pull high and then pull low  $\overline{\text{EN}}$  to release the fault condition.

### Selecting the Input and Output Capacitors

In most applications, the most important is the high frequency decoupling capacitor on the input of the RT9503A.

A 1 $\mu$ F ceramic capacitor, placed in close proximity to input

pin and GND pin is recommended. In some applications depending on the power supply characteristics and cable length, it may be necessary to add an additional  $10\mu$ F ceramic capacitor to the input. The RT9503A requires a small output capacitor for loop stability. A  $1\mu$ F ceramic capacitor placed between the BATT pin and GND is typically sufficient.

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

 $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = \left(\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}\right) / \theta_{\mathsf{J}\mathsf{A}}$ 

Where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications of RT9503A, the maximum junction temperature is 125°C and T<sub>A</sub> is the ambient temperature. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For WQFN-16L 3x3 packages, the thermal resistance,  $\theta_{JA}$ , is 68°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at T<sub>A</sub> = 25°C can be calculated by the following formula :

 $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})}$  = (125°C - 25°C) / (68°C/W) = 1.471W for WQFN-16L 3x3 package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . For RT9503A package, the derating curve in Figure 7 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

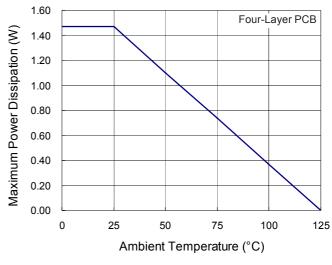


Figure 7. Derating Curves for RT9503A Package

### Layout Consideration

The RT9503A is a fully integrated low cost single-cell Lilon battery charger ideal for portable applications. Careful PCB layout is necessary. The following guidelines should be strictly followed to achieve best performance of the RT9503A.

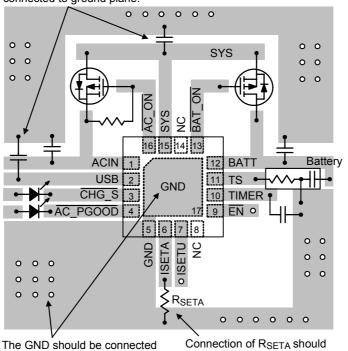
- Input capacitor should be placed close to the IC and connected to the ground plane. The trace of the input in the PCB should be placed far away from the sensitive devices or shielded by the ground.
- The GND should be connected to a strong ground plane for heat sinking and noise protection.
- Connection of the R<sub>SETA</sub> should be isolated from other noisy traces. A short wire is recommended to prevent EMI and noise coupling.
- Output capacitor should be placed close to the IC and connected to ground plane to reduce noise coupling.

# RT9503A





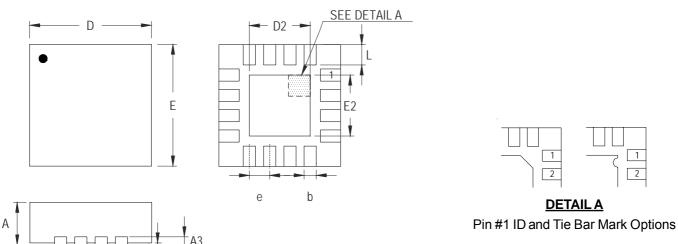
The capacitors should be placed close to the IC and connected to ground plane.



The GND should be connected to a strong ground plane for heat sinking and noise protection. Connection of R<sub>SETA</sub> should be isolated from other noisy traces.

Figure 8. PCB Layout Guide

### **Outline Dimension**



Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.180	0.300	0.007	0.012	
D	2.950	3.050	0.116	0.120	
D2	1.300	1.750	0.051	0.069	
E	2.950	3.050	0.116	0.120	
E2	1.300	1.750	0.051	0.069	
е	0.500		0.0	)20	
L	0.350	0.450	0.014	0.018	

W-Type 16L QFN 3x3 Package

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