

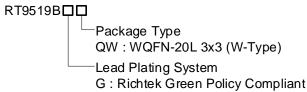
Linear Single Cell Li-lon Battery Charger with Auto Power Path Management

General Description

The RT9519B is an integrated single-cell Li-ion battery charger with auto power path management (APPM). No external MOSFETs are required. The RT9519B enters sleep mode when power is removed. Charging tasks are optimized by using a control algorithm to vary the charge rate, including pre-charge mode, fast charge mode and constant voltage mode. For the RT9519B, the charge current can also be programmed with an external resistor and modified with an external GPIO. The scope that the battery regulation voltage can be modified with an external GPIO depends on the battery temperature. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. The charging task will always be terminated in constant voltage mode when the charging current reduces to the termination current of 10% x ICHG FAST. Other features include under voltage protection and over voltage protection for VIN supply.

The recommended junction temperature range spans from -40°C to 125°C, while the ambient temperature range extends from -40°C to 85°C.

Ordering Information



Note:

Richtek products are Richtek Green Policy compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.

Marking Information



46=: Product Code YMDNN: Date Code

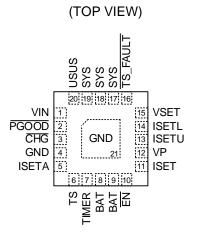
Features

- 28V Maximum Rating for VIN Power
- Selectable Power Current Limit (0.1A / 0.5A / 1.5A)
- Integrated Power MOSFETs
- Auto Power Path Management (APPM)
- Battery Charging Current Control
- Battery Regulation Voltage Control
- Programmable Charging Current and Safe Charge Timer
- Under Voltage Protection, Over Voltage Protection
- Power Good and Charge Status Indicator
- Optimized Charge Rate via Thermal Feedback
- Thin 20-Lead WQFN Package
- RoHS Compliant and Halogen Free

Applications

- Digital Cameras
- PDAs and Smart Phones
- Portable Instruments

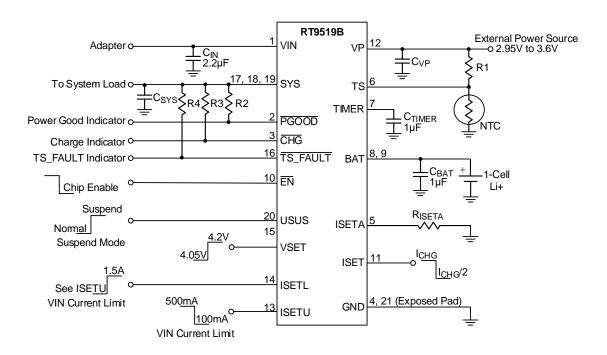
Pin Configuration



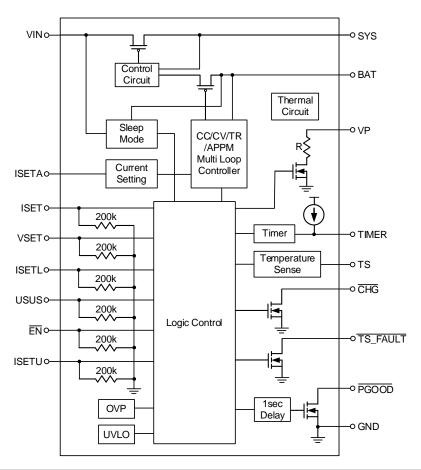
WQFN-20L 3x3



Typical Application Circuit



Functional Block Diagram





Functional Pin Description

Pin No.	Pin Name	Pin Function
1	VIN	Supply voltage input.
2	PGOOD	Power good status output. Active low, open-drain output.
3	CHG	Charger status output. Active low, open-drain output.
4, 21 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
5	ISETA	Charge current set input. Connect a resistor (RISETA) between ISETA and GND.
6	TS	Temperature Sense Input. The TS pin connects to a battery's thermistor to determine if the battery is too hot or too cold to charge. If the battery's temperature is out of range, charging is paused until it re-enters the valid range. TS also detect battery (with NTC) is present or not.
7	TIMER	Safe charge timer setting.
8, 9	BAT	Battery charge current output.
10	ĒN	Charge enable. Active high input. 200k Ω pull low.
11	ISET	Half charge current set input. Control by external GPIO, L = ICHG1 / 2, H = ICHG1, $200k\Omega$ pull low.
12	VP	This pin must be provided a regulated voltage from 2.95V to 3.6V by external power.
13	ISETU	VIN current limit control input. When $\overline{\text{ISETL}}$ = H, L = 100mA, H = 500mA, 200k Ω pull low.
14	ISETL	VIN current limit control input. H : see ISETU, L = 1.5A, $200k\Omega$ pull low.
15	VSET	Battery regulation set input. Control by external GPIO. L = 4.05V, H = 4.2V, $200k\Omega$ pull low.
16	TS_FAULT	TS fault status output. Open-drain output. If TS is of fault (too cold, too hot) or VP triggers UVLO, TS_FAULT = L. If TS and VP are normal, TS_FAULT = H.
17, 18, 19	SYS	System connect pin. Connect this pin to system with a minimum $10\mu\text{F}$ ceramic capacitor connected to GND.
20	USUS	VIN suspend control input. H = Suspend, L = No suspend. $200k\Omega$ pull low.



Absolute Maximum Ratings (Note 1)

• Supply Voltage, VIN	0.3V to 28V
• CHG, PGOOD	0.3V to 28V
• Other Pins	0.3V to 6V
CHG, PGOOD Continuous Current	- 20mA
BAT Continuous Current (total in two pins) (Note 2)	- 2.5A
 Power Dissipation, PD @ TA = 25°C 	
WQFN 20L 3x3	- 1.471W
Package Thermal Resistance (Note 2)	
WQFN 20L 3x3, θ JA	- 68°C/W
WQFN 20L 3x3, θJC	- 7.5°C/W
Lead Temperature (Soldering, 10sec.)	- 260°C
Junction Temperature	- 150°C
Storage Temperature Range	65°C to 150°C
ESD Susceptibility (Note 3)	
HBM (Human Body Model)	- 2kV
MM (Machine Model)	- 200V
Recommended Operating Conditions (Note 4)	
Supply Input Voltage Range, VIN (ISETL = L)	- 4.4V to 6V
Supply Input Voltage Range, VIN (ISETL = H)	- 4.5V to 6V

Electrical Characteristics

(V_{IN} = 5V, V_{BAT} = 4V, T_A = 25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Supply Input						
VIN Under Voltage Lockout Threshold	Vuvlo	VIN = 0V to 4V	3.1	3.3	3.5	V
VIN Under Voltage Lockout Hysteresis	ΔVυνιο	V _{IN} = 4V to 0V		240		mV
VIN Supply Current	loupply	$ISYS = IBAT = 0mA, \overline{EN} = L$ (VBAT > VREGx)		1	2	mA
VIN Supply Current	ISUPPLY	$I_{SYS} = I_{BAT} = 0mA, \overline{EN} = H$ (VBAT > VREGx)		0.8	1.5	mA
VIN Suspend Current	lusus	V _{IN} = 5V, USUS = H		195	300	μΑ
VBAT Sleep Leakage Current	ISLEEP	VBAT > VIN (VIN = 0V)		5	15	μΑ
VIN-BAT VOS Rising	Vos_H			200	300	mV
VIN-BAT VOS Falling	Vos_L		10	50		mV



Parameter		Symbol	Test Conditions	Min	Тур	Max	Unit
Voltage Regulation		•				l e	
System Regulation Volt	tage	Vsys	Isys = 800mA	4.3	4.4	4.5	V
Battery Regulation Voltage		VREG1	Loading = 20mA When VSET = H	4.16	4.2	4.23	V
Battery Regulation Volt	age	VREG2	Loading = 20mA When VSET = L	4.01	4.05	4.08	٧
APPM Regulation Volta	ige	ΔVΑΡΡΜ	VSYS - ΔVAPPM	120	200	280	mV
DPM Regulation Voltage	je	VDPM	ISETL = L	4.3	4.4	4.5	V
VIN to VSYS MOSFET	Ron	RDS(ON)	IVIN = 1000mA		0.2	0.35	Ω
BAT to VSYS MOSFET	Ron	RDS(ON)	VBAT = 4.2V, ISYS = 1A		0.05	0.1	Ω
Re-Charge Threshold		ΔVREGCHG	Battery Regulation – Recharge level	60	100	140	mV
Current Regulation							
ISETA Set Voltage (Fas Phase)	st Charge	VISETA	$V_{BAT} = 4V$, $R_{ISETA} = 1k\Omega$		2		V
VIN Charge Setting Ra	nge	ICHG		100		1200	mA
VIN Charge Current Accuracy1		ICHG1	VBAT = 4V, RISETA = $1k\Omega$ ISET = H	570	600	630	mA
VIN Charge Current Accuracy2		ICHG2	VBAT = 3.8V, RISETA = $1k\Omega$ ISET = L	285	300	315	mA
			ISETL = L (1.5A Mode)	1	1.5	1.8	Α
VIN Current Limit		IVIN	ISETL = H, ISETU = H (500mA Mode)	430	475	500	mA
			ISETL = H, ISETU = L (100mA Mode)	70	95	100	mA
Pre-Charge							
BAT Pre-Charge Thresl	hold	VPRECH	BAT Falling	2.7	2.8	2.9	V
BAT Pre-Charge Thresh Hysteresis	hold	ΔVPRECH			200		mV
Pre-Charge Current		ICHG_PRE	VBAT = 2V	5	10	15	%
Charge Termination D	etection						
Termination Current Ra Charge (Except USB10		ITERM	ISETL = L, ISETU = H ISETL = H, ISETU = X	5	10	15	%
Termination Current Ratio to Fast Charge (USB100 Mode)		ITERM2	ISETL = L, ISETU = L		3.3		%
Login Input/Output		•		•	•		
CHG Pull Down Voltage	9	VCHG	I CHG = 5mA		200		mV
PGOOD Pull Down Voltage		VPGOOD	I PGOOD = 5mA		200		mV
TS_FAULT Pull Down \	√oltage	VTS_FAULT	ITS_FAULT = 5mA		200		mV
EN, ISETL, USUS, ISETU, VSET, ISET	Logic-High	VIH		1.5			V
Threshold Voltage	Logic-Low	VIL				0.4	

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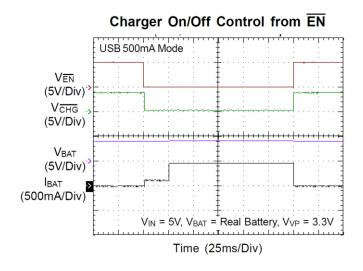
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Protection	1		l			
Thermal Regulation	T _{REG}			125		°C
Thermal Shutdown Temperature	TsD			155		°C
Thermal Shutdown Hysteresis	ΔTsd			20		°C
Over Voltage Protection	Vovp	V _{IN} Rising	6.25	6.5	6.75	V
Over Voltage Protection Hysteresis	ΔVOVP	VIN = 7V to 5V, $VOVP - \Delta VOVP$		100	1	mV
Output Short Circuit Detection Threshold	VSHORT	VBAT – VSYS		300		mV
Time						
Pre-Charge Fault Time	tpchg	CTIMER = 1μ F (1/8 x tFCHG)	1440	1800	2160	S
Fast Charge Fault Time	tFCHG	CTIMER = 1μF	11520	14400	17280	S
PGOOD Deglitch Time	tPGOOD	Time measured from VIN : 0→5V 1μs rise time to PGOOD = L		1	1	S
Input Over Voltage Blanking Time	tovp			50	1	μS
Pre-Charge to Fast-Charge Deglitch Time	tpF			25	1	ms
Fast-Charge to Pre-Charge Deglitch Time	tFP			25	-	ms
Termination Deglitch Time	ttermi			25		ms
Recharge Deglitch Time	trechg			100	1	ms
Input Power Loss to SYS LDO Turn-Off Delay Time	t _{NO_IN}			25	1	ms
Short Circuit ,Deglitch Time	tshort			250	1	μS
Short Circuit Recovery Time	tshort-r			64		ms
Other	•					
VP (External used only)	VVP		2.95		3.6	V
VP Under Voltage Lockout Threshold		Falling Threshold		0.8		V
TS Battery Detect Threshold	VTS		2.75	2.85	2.95	V
NTC	•					
Low Temperature Trip Point	VCOLD	Rising Threshold	73	74	75	% of VP
Low Temperature Trip Point Hysteresis	ΔVCOLD			1		% of VP
High Temperature Trip Point	Vнот	Falling Threshold	27	28	29	% of VP
High Temperature Trip Point Hysteresis	ΔVΗΟΤ			1		% of VP

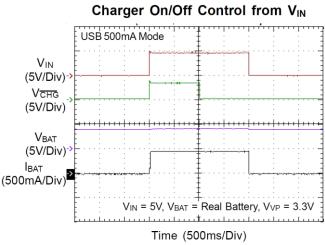


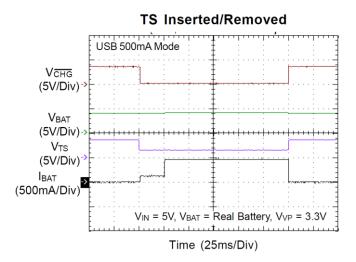
- Note 1. Stresses beyond those listed under "Absolute Maximum Ratings" June cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions June affect device reliability.
- Note 2. Guaranteed by design.
- Note 3. θJA is measured under natural convection (still air) at TA = 25°C with the component mounted on a high effective-thermalconductivity four-layer test board on a JEDEC 51-7 thermal measurement standard. HJC is measured at the exposed pad of the package.
- Note 4. Devices are ESD sensitive. Handling precautions are recommended.
- Note 5. The device is not guaranteed to function outside its operating conditions.

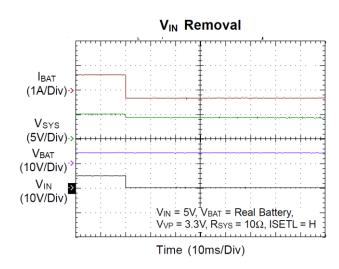


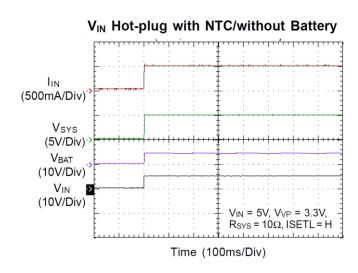
Typical Operating Characteristics

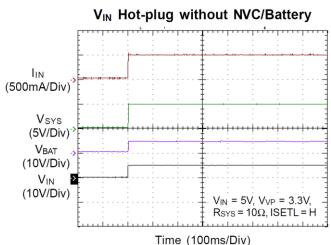




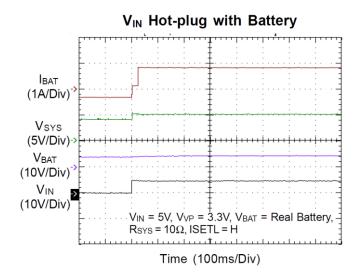


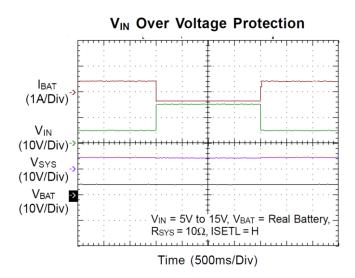


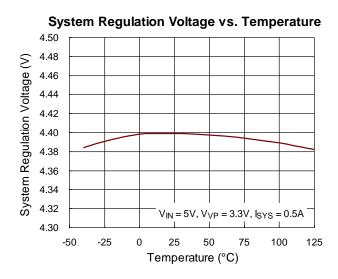


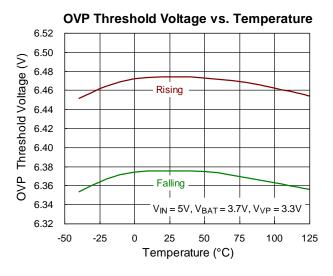


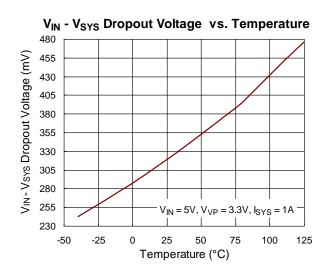


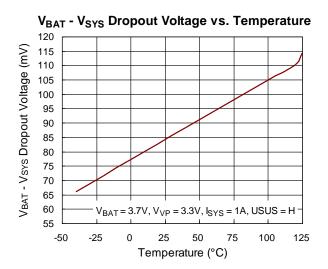






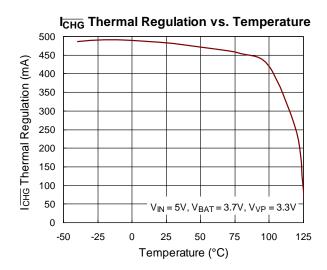


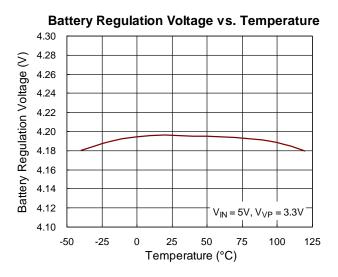


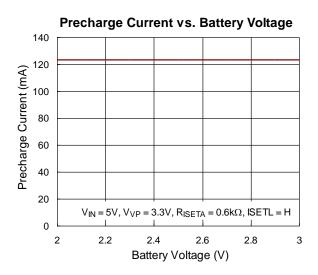


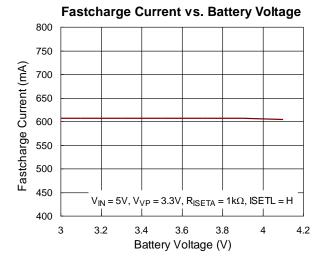
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Application Information

Richtek's component specification does not include the following information in the Application Information section. Thereby no warranty is given regarding its validity and accuracy. Customers should take responsibility to verify their own designs and reserve suitable design margin to ensure the functional suitability of their components and systems.

The RT9519B is a fully integrated single-cell Li-ion battery charger ideal for portable applications. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. Other features include under voltage protection and over voltage protection.

$$I_{CHG_FAST} = \frac{V_{ISETA}}{R_{ISETA}} \times 150$$

ICHG_PRE = 10% x ICHG_FAST

Pre-charge Mode

When the output voltage is lower than 2.8V, the charging current will be reduced to a fast-charge current ratio set by RISETA to protect the battery life time.

Fast-charge Mode

When the output voltage is higher than 3V, the charging current will be equal to the fast-charge current set by RISETA.

Constant-Voltage Mode

When the output voltage is near 4.2V, and the charging current fall below the termination current, after a deglitch time check of 25ms, the charger will become disabled and CHG will go from L to H.

Re-charge Mode

When the chip is in charge termination mode, the charging current will gradually go down to zero. However, once the voltage of the battery drops to below 4.1V, there will be a deglitch time of 100ms and then the charging current will resume again.

Charging Current Decision

The charge current can be set according to the following equations:

$$I_{CHG_FAST} = \frac{V_{ISETA}}{R_{ISETA}} \times 300$$

If ISET = L (for ICHG2)

Time Fault

During the fast charge phase, several events may increase the charging time.

For example the system load current may have activated the APPM loop which reduces the available charging current, the device has entered thermal regulation because the IC junction temperature has exceeded TREG. During each of these events, if 3V < VBAT < 4.1V, the internal charging time is slowed down proportionately to the reduction in charging current. However, once the duration exceeds the fault time, the CHG output will flash at approximately 2Hz to indicate a fault condition and the charge current will be reduced to about 1mA.

$$t_{\text{FCHG_true}} = t_{\text{FCHG}} \times \frac{2V}{V_{\text{ISETA}}}$$

tFCHG true: modified timer in fast

tFCHG: original timer in fast charger

$$t_{\text{FCHG}} = 14400 \times \left(\frac{C_{\text{TIMER}}}{1 \mu F}\right)$$

$$t_{PCHG} = \frac{t_{FCHG}}{8}$$

tpchg: timer in pre-charge

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Time fault release methods:

- (1) Re-plug power
- (2) Toggle EN
- (3) Enter/exit suspend mode
- (4) Remove Battery
- (5) OVP

Note that the fast charge fault time is independent of the charge current.

Power Good

VIN Power Good (PGOOD = L)

Input State	PGOOD Output
VIN < VUVLO	High Impedance
VUVLO < VIN < VBAT + VOS_H	High Impedance
V _{BAT} + V _{OS} _H < V _{IN} < V _{OVP}	Low Impedance
VIN > VOVP	High Impedance

Charge State Indicator

Charge State	CHG Output
Charging	Low
Charging Suspended by Thermal Loop	(for first charge cycle)
Safety Timers Expired	2Hz Flash
Charging Done	
Recharging after Termination	High Impedance
IC Disabled or no Valid Input Power	

Battery Pack Temperature Monitoring

The battery pack temperature monitoring function can be realized by connecting the TS pin to an external Negative Temperature Coefficient (NTC) thermistor to prevent over temperature condition. Charging is suspended when the voltage at the TS pin is out of normal operating range. The internal timer is then paused, but the value is maintained. When the TS pin voltage returns back to normal operating range, charging will resume and the safe charge timer will continue to count down from the point where it was suspended. Note that although charging is suspended due to the battery pack temperature fault, the CHG pin will continue to remain low and indicate charging.

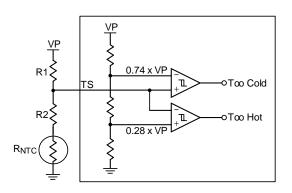


Figure 1

Too Cold Temperature

RCOLD = RNTC

Too Hot Temperature

RHOT = RNTC

$$\frac{R2 + R_{COLD}}{R_{COLD} + R1 + R2} = 0.74 \tag{1}$$

$$\frac{R2 + R_{HOT}}{R_{HOT} + R1 + R2} = 0.28 \tag{2}$$

$$R1 = \frac{R_{COLD} - R_{HOT}}{2.457}$$

$$R2 = 0.359 \times R1 - R_{HOT}$$

If R2 < 0

$$\frac{R_{COLD}}{R_{COLD} + R1} = 0.74 \tag{3}$$

From (3)

$$R1 = \frac{R_{COLD}}{0.74} - R_{COLD}$$

Charge Enable

When EN is High, the charger turns on. When EN is low, the charger turns off. $\overline{\text{EN}}$ is pulled low at the initial condition.

VIN Input Current Limit

ISETL	ISETU	VIN Input Current Limit
L	L	95mA
L	Н	475mA
Н	Х	1.5A



Suspend Mode

Set USUS = H, and the charge will enter Suspend Mode. In the Suspend Mode, CHG is in high impedance and IUSUS(MAX) $< 300 \mu A$.

Power Switch

For the RT9519B, there are three power scenarios:

(1) When a battery and an external power supply (USB or adapter) are connected simultaneously:

If the system load requirements exceed that of the input current limit, the battery will be used to supplement the current to the load. However, if the system load requirements are less than that of the input current limit, the excess power from the external power supply will be used to charge the battery.

- (2) When only the battery is connected to the system: The battery provides the power to the system.
- (3) When only an external power supply is connected to the system:

The external power supply provides the power to the system.

Input DPM Mode

For the RT9519B, the input voltage is monitored when the USB100 or USB500 is selected. If the input voltage is lower than VDPM, the input current limit will be reduced to stop the input voltage from dropping any further. This can prevent the IC from damaging improperly configured or inadequately designed USB sources.

APPM Mode

Once the sum of the charging and system load currents becomes higher than the maximum input current limit, the SYS pin voltage will be reduced. When the SYS pin voltage is reduced to VAPPM, the RT9519B will automatically operate in APPM mode. In this mode, the charging current is reduced while the SYS current is

increased to maintain system output. In APPM mode, the battery termination function is disabled.

Battery Supplement Mode Short Circuit Protect

In APPM mode, the SYS voltage will continue to drop if the charge current is zero and the system load increases beyond the input current limit. When the SYS voltage decreases below the battery voltage, the battery will kick in to supplement the system load until the SYS voltage rises above the battery voltage.

While in supplement mode, there is no battery supplement current regulation. However, a built in short circuit protection feature is available to prevent any abnormal current situations. While the battery is supplementing the load, if the difference between the battery and SYS voltage becomes more than the short circuit threshold voltage, SYS will be disabled. After a short circuit recovery time, tSHORT R, the counter will be restarted. In supplement mode, the battery termination function is disabled. Note that for the battery supply mode exit condition, VBAT - VSYS < 0V.

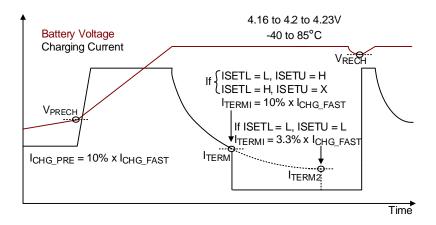
Thermal Regulation and Thermal Shutdown

The RT9519B provides a thermal regulation loop function to monitor the device temperature. If the die temperature rises above the regulation temperature, TREG, the charge current will automatically be reduced to lower the die temperature. However, in certain circumstances (such as high V_{IN}, heavy system load, etc.) even with the thermal loop in place, the die temperature may still continue to increase. In this case, if the temperature rises above the thermal shutdown threshold, TSD, the internal switch between VIN and SYS will be turned off. The switch between the battery and SYS will remain on, however, to allow continuous battery power to the load. Once the die temperature decreases by ΔT_{SD} , the internal switch between VIN and SYS will be turned on again and the device returns to normal thermal regulation.

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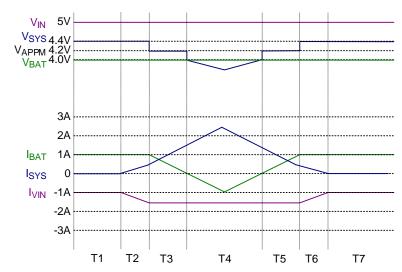


Charging Profile



APPM Profile

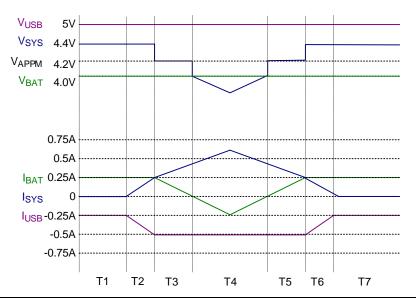
1.5A Mode:



	Isys	Vsys	Ivin	Іват
T1, T7	0	SYS Regulation Voltage	CHG_MAX	CHG_MAX
T2, T6	< IVIN_OC- CHG_MAX	SYS Regulation Voltage	Isys + CHG_MAX	CHG_MAX
T3, T5	> IVIN_OC- CHG_MAX < IVIN_OC	Auto Charge Voltage Threshold	VIN_OC	VIN_OC-ISYS
T4	> IVIN_OC	VBAT-IBAT X RDS(ON)	VIN_OC	ISYS-IVIN_OC



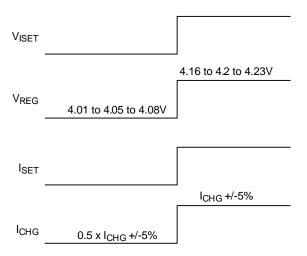
USB 500mA Mode:



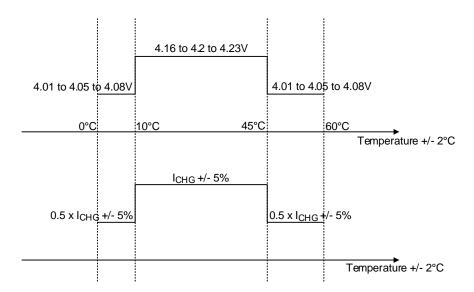
	Isys	Isys Vsys		I BAT
T1, T7	0	SYS Regulation Voltage	CHG_MAX	CHG_MAX
T2, T6	< IVIN_OC (USB)- CHG_MAX	SYS Regulation Voltage	Isys + CHG_MAX	CHG_MAX
T3, T5	> IVIN_OC (USB)- CHG_MAX < IVIN_OC (USB)	Auto Charge Voltage Threshold	IVIN_OC (USB)	IVIN_OC (USB)-Isys
T4	> IVIN_OC (USB)	VBAT-IBAT x RDS(ON)	IVIN_OC (USB)	Isys-Ivin_oc (USB)



VSET vs. VREG, ISET vs. ICHG



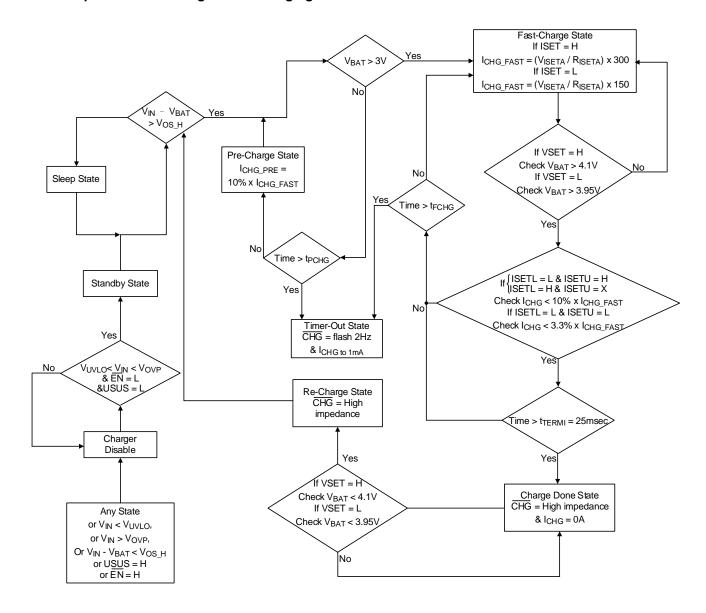
For JEITA Battery Temperature Standard: CV regulation voltage will change at the following battery Temp ranges 0°C to 10°C and 45°C to 60°C CC regulation current will change at the following battery Temp ranges 0°C to 10°C and 45°C to 60°C



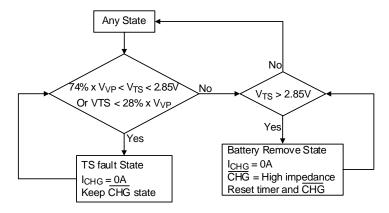
October 2023



RT9519B Operation State Diagram for Charging



Operation State Diagram for TS PIN



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Thermal Considerations

The junction temperature should never exceed the absolute maximum junction temperature T_J(MAX), listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula:

 $PD(MAX) = (TJ(MAX) - TA) / \theta JA$

where $T_{J(MAX)}$ is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance, θ_{JA} , is highly package dependent. For a WQFN-20L 3x3 package, the thermal resistance, θ_{JA} , is 68°C/W on a standard JEDEC 51-7 high effective-thermal-conductivity four-layer test board. The maximum power dissipation at $T_A = 25$ °C can be calculated as below :

 $P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (68^{\circ}C/W) = 1.471W$ for a WQFN-20L 3x3 package.

The maximum power dissipation depends on the operating ambient temperature for the fixed $T_{J(MAX)}$ and the thermal resistance, θ_{JA} . The derating curves in Figure 2 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

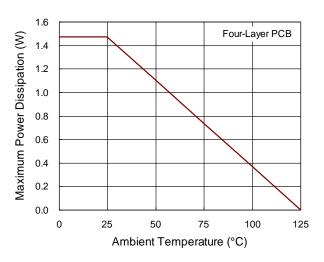


Figure 2. Derating Curves for RT9519B Package

Layout Considerations

The RT9519B is a fully integrated low cost single cell Liion battery charger ideal for portable applications. Careful PCB layout is necessary. For best performance, place all peripheral components as close to the IC as possible.

A short connection is highly recommended. The following guidelines should be strictly followed when designing a

PCB layout for the RT9519B.

- ▶ Input and output capacitor should be placed close to the IC and connected to ground plane. The input trace on the PCB should be placed far away from sensitive devices and shielded by the ground.
- ▶ The GND and exposed pad should be connected to a strong ground plane for heat sinking and noise protection.
- ► The connection of RISETA should be isolated from other noisy traces. A short wire is recommended to prevent EMI and noise coupling.

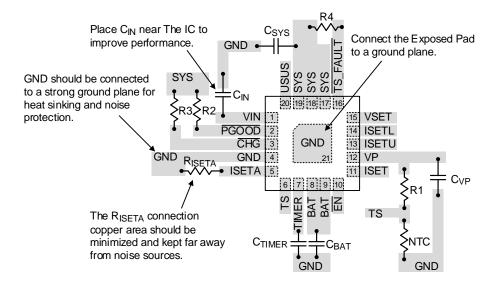
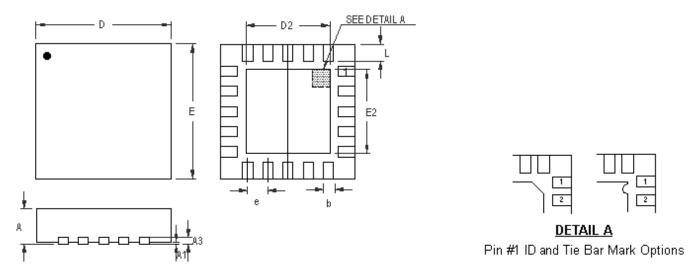


Figure 3. PCB Layout Guide



Outline Dimension



Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

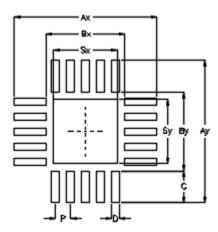
Cymhal	Dimensions I	n Millimeters	Dimensions In Inches			
Symbol	Min	Max	Min	Max		
А	0.700	0.800	0.028	0.031		
A1	0.000	0.050	0.000	0.002		
A3	0.175	0.250 0.007		0.010		
b	0.150	0.250	0.006	0.010		
D	2.900	3.100	0.114	0.122		
D2	1.650	1.750	0.065	0.069		
E	2.900	3.100	0.114	0.122		
E2	1.650	1.750	0.065	0.069		
е	0.4	100	0.0)16		
L	0.350	0.450	0.014	0.018		

W-Type 20L QFN 3x3 Package

DS9519B-03



Footprint Information

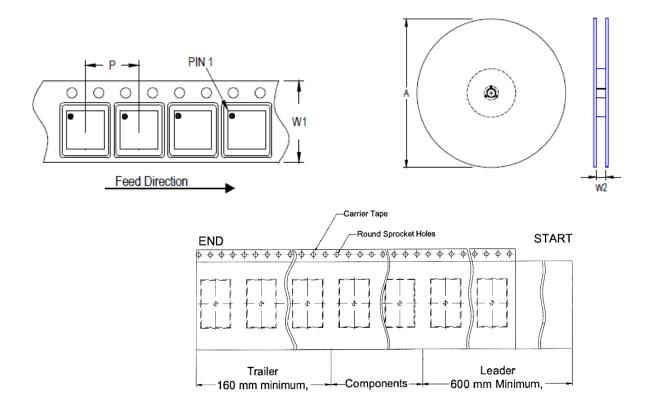


Doolsogo	Number of	Footprint Dimension (mm)							Toloropoo		
Package	Pin	Р	Ax	Ay	Вх	Ву	С	D	Sx	Sy	Tolerance
V/W/U/XQFN3*3-20	20	0.40	3.80	3.80	2.10	2.10	0.85	0.20	1.70	1.70	±0.05

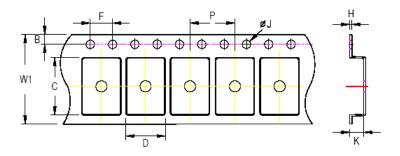


Packing Information

Tape and Reel Data



Package Type	Tape Size (W1) (mm)	Pocket Pitch (P) (mm)	Reel Si	Reel Size (A) (mm) (in)		Trailer (mm)	Leader (mm)	Reel Width (W2) Min./Max. (mm)
QFN/DFN 3x3	12	8	180	7	1,500	160	600	12.4/14.4



- C, D and K are determined by component size.

 The clearance between the components and the cavity is as follows:
- For 12mm carrier tape: 0.5mm max.

Tape Size W1		Р		В		F		Ø٦		Н
Tape Size	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Max.
12mm	12.3mm	7.9mm	8.1mm	1.65mm	1.85mm	3.9mm	4.1mm	1.5mm	1.6mm	0.6mm



Tape and Reel Packing

Step	Photo/Description	Step	Photo/Description
1	Reel 7"	4	RICHTER OF THE PROPERTY OF THE
2	Management of the state of the	5	
3	HIC & Desiccant (1 Unit) inside Caution label is on backside of Al bag	6	Outer box Carton A

Container	R	eel		Box			Carton			
Package	Size	Units	Item	Size(cm)	Reels	Units	Item	Size(cm)	Boxes	Unit
OFN 8 DEN 2v2	QFN & DFN 3x3 7" 1,500 Box A Box E		Box A	18.3*18.3*8.0	3	4,500	Carton A	38.3*27.2*38.3	12	54,000
QFN & DFN 3X3			Box E	18.6*18.6*3.5	1	1,500	For Combined or Partial Reel.			

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Packing Material Anti-ESD Property

Surface Resistance	Aluminum Bag	Reel	Cover tape	Carrier tape	Tube	Protection Band
Ω /cm 2	10 ⁴ to 10 ¹¹					

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Datasheet Revision History

Version	Date	Description	Item
02	2022/11/23	Modify	Electrical Characteristics on P5 Application Information on P11 Footprint Information on P21 Packing Information on P22, 23, 24
03	2023/10/2	Modify	General Description on P1 Ordering Information on P1 Electrical Characteristics on P5 Application Information on P11