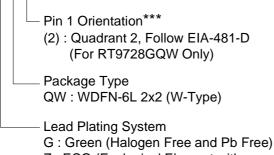
## 120mΩ, 1.3A Power Switch with Programmable Current Limit

### **General Description**

The RT9728 is a cost effective, low voltage, single P-MOSFET high side power switch IC for USB application with a programmable current limit feature. Low switch-on resistance (typ.120m $\Omega$ ) and low supply current (typ. 120µA) are realized in this IC. The RT9728 offers a programmable current limit threshold between 75mA and 1.3A (typ.) via an external resistor. The ±10% current limit accuracy can be realized for all current limit settings. Besides, a flag output is available to indicate fault conditions to the local USB controller. Furthermore, the chip also integrates an embedded delay function to prevent mis-operation from happening due to high inrush current. The RT9728 is an ideal solution for USB power supply and can support flexible applications since it is functional for various current limit requirements. It is available in WDFN-6L 2x2 package.

### **Ordering Information**

### RT9728



- Z : ECO (Ecological Element with
- Halogen Free and Pb free)

Note :

\*\*\*Empty means Pin1 orientation is Quadrant 1

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

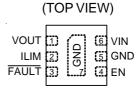
### **Features**

- ±10% Current Limit Accuracy @ 1.3A
- Adjustable Current Limit : 75mA to 1.3A (typ.)
- Meet USB Current Limiting Requirements
- Operating Voltage Range : 2.5V to 5.5V
- Reverse Input Output Voltage Protection
- Built-in Soft-Start
- 120mΩ High Side MOSFET
- 120µA Supply Current
- 15-kV ESD Protection per IEC 61000-4-2 (With **External Capacitance)**
- Nemko Approved IEC62368-1
- RoHS Compliant and Halogen Free

### Applications

- USB Bus/Self Powered Hubs
- USB Peripheral Ports
- ACPI Power Distribution
- Battery Power Equipment
- 3G/3.5G Data Card, Set-Top Boxes

## Pin Configuration



WDFN-6L 2x2

## **Marking Information**

#### **RT9728GQW**



W: Date code

### **RT9728ZQW**

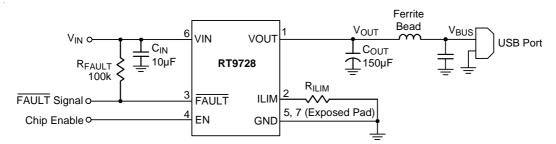
J0W

•  J0 : Product Code W : Date Code

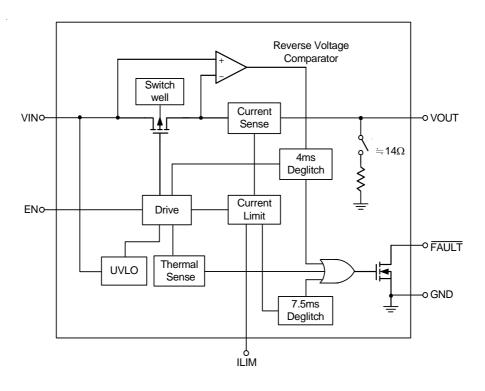
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## **Typical Application Circuit**



## **Functional Block Diagram**



### **Functional Pin Description**

Pin No.	Pin Name	Pin Function
1	VOUT	Power switch output.
2	ILIM	External resistor is used to set current limit threshold, and 19.1k $\Omega \leq R_{ILIM} \leq 232 k \Omega$ is recommended.
3	FAULT	Active-low open-drain output. Asserted during over-current, over-temperature, or reverse voltage conditions.
4	EN	Enable control input.
5, 7 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
6	VIN	Supply input voltage.

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## Absolute Maximum Ratings (Note 1)

<ul> <li>Supply Input Voltage</li> <li>Other Pin Voltage</li> </ul>	
<ul> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> <li>WDFN-6L 2x2</li> </ul>	0.606W
Package Thermal Resistance (Note 2)	
WDFN-6L 2x2, $\theta_{JA}$	165°C/W
WDFN-6L 2x2, $\theta_{JC}$	7°C/W
Lead Temperature (Soldering, 10 sec.)	260°C
Junction Temperature	150°C
Storage Temperature Range	$-65^{\circ}C$ to $150^{\circ}C$
ESD Susceptibility (Note 3)	
HBM (Human Body Model)	2kV

## Recommended Operating Conditions (Note 4)

Supply Input Voltage, VIN	- 2.5V to 5.5V
Junction Temperature Range	- –40°C to 125°C
Ambient Temperature Range	40°C to 85°C

### **Electrical Characteristics**

(V<sub>IN</sub> = 3.6V, 19.1k $\Omega$   $\leq$  R<sub>ILIM</sub>  $\leq$  232k $\Omega$ , T<sub>A</sub> = 25°C, unless otherwise specified)

Parameter		Symbol	Test Conditions		Min	Тур	Мах	Unit	
EN Threshold	Logic-High	Viн			1.1			V	
Voltage	Logic- Low	VIL					0.66	V	
Current-Limit Threshold Resistor Range		RILIM			19.1		232	kΩ	
Under Voltage L	_ockout		V <sub>IN</sub> rising			2.35			
Threshold		UVLO	V <sub>IN</sub> falling			2.10		V	
Shutdown Current		I <sub>SHDN</sub>	$V_{IN} = 5.5V$ , no load on $V_{OUT}$ , $V_{EN} = 0V$			1	3	μA	
Quiescent Current		IQ	$V_{IN} = 5.5V$ , no load on $V_{OUT}$	$R_{ILIM} = 20k\Omega$		120	170	μΑ	
				R <sub>ILIM</sub> = 210kΩ		120	170		
Reverse Leakage Current		I <sub>REV</sub>	V <sub>OUT</sub> = 5.5V, V <sub>IN</sub> = 0V			1	3	μA	
Static Drain-Source On-State Resistance		R <sub>DS(ON)</sub>	$I_{SW} = 0.2A$			120		mΩ	
Current Limit			$R_{ILIM} = 20k\Omega$		1166	1295	1425	· mA	
		I <sub>LIM</sub>	R <sub>ILIM</sub> = 49.9kΩ		468	520	572		
			R <sub>ILIM</sub> = 210kΩ		110	130	150		
			ILIM shorted to VIN		50	75	100		
Reverse Voltage Comparator Trip Point (V <sub>OUT</sub> – V <sub>IN</sub> )						135		mV	

## **RT9728**



Parameter	Symbol	Test Conditions		Тур	Max	Unit
FAULT Output Low Voltage	Vol	I <sub>FAULT</sub> = 1mA		180		mV
FAULT Off State Leakage		VFAULT = 5.5V	-	1	-	μA
FAULT Deglitch		FAULT assertion or de-assertion due to over current condition	5	7.5	10	
		FAULT assertion or de-assertion due to reverse voltage condition	2	4	6	ms
FAULT Flag Assertion Offset	$V_{\overline{FAULT}\_OFS}$	Offset between fault flag assertion level versus ILIM trigger level (Note 5)	-100		0	mA
Thermal Shutdown Temperature	T <sub>SD</sub>	(Note 5)		160		°C

**Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

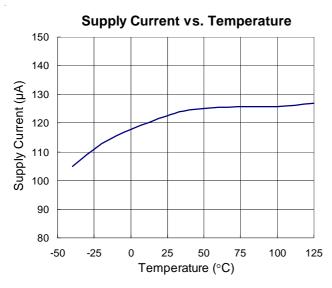
Note 2.  $\theta_{JA}$  is measured at  $T_A = 25^{\circ}C$  on a low effective thermal conductivity single-layer test board per JEDEC 51-3.  $\theta_{JC}$  is measured at the exposed pad of the package.

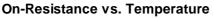
Note 3. Devices are ESD sensitive. Handling precautions are recommended.

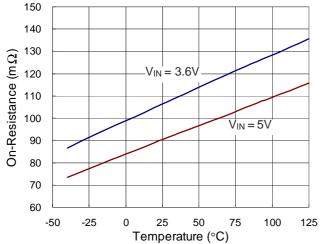
- Note 4. The device is not guaranteed to function outside its operating conditions.
- Note 5. Guarantee by design.

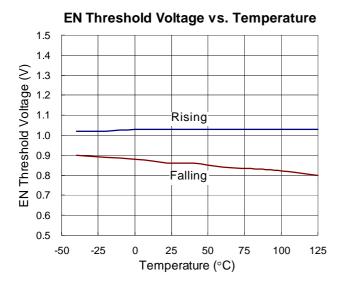
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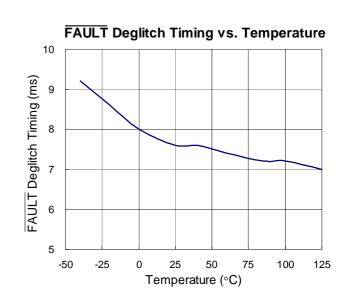
## **Typical Operating Characteristics**



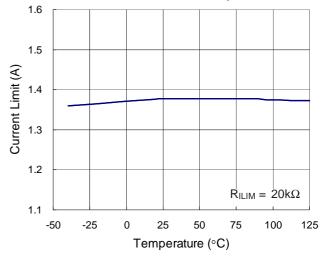




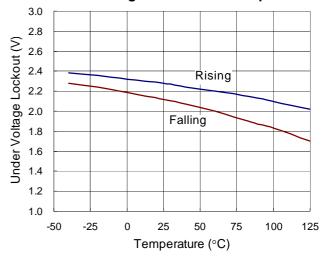




Current Limit vs. Temperature



Under Voltage Lockout vs. Temperature

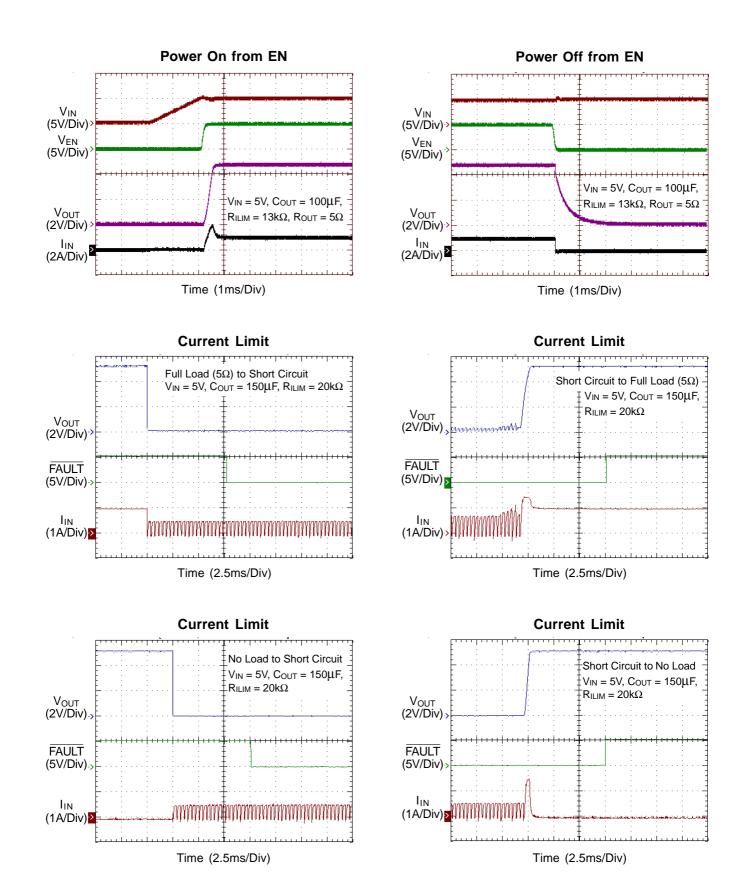


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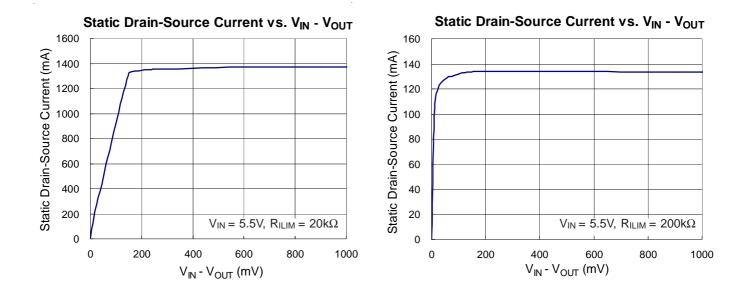
DS9728-10 February 2023

## **RT9728**





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## **Application Information**

Richtek's component specification does not include the following information in the Application Information section. Thereby no warranty is given regarding its validity and accuracy. Customers should take responsibility to verify their own designs and to ensure the functional suitability of their components and systems.

The RT9728 is a single P-MOSFET high side power switch with active high enable input, optimized for self powered and bus powered Universal Serial Bus (USB) applications. The switch's low  $R_{DS(ON)}$  meets USB voltage drop requirements and a flag output is available to indicate fault conditions to the local USB controller.

#### **Current Limiting and Short Circuit Protection**

When a heavy load or short circuit situation occurs while the switch is enabled, large transient current may flow through the device. The RT9728 includes a current limit circuitry to prevent these large currents from damaging the MOSFET switch and the hub downstream ports. The RT9728 provides an adjustable current limit threshold between 120mA and 1.3A (typ.) via an external resistor,  $R_{ILIM}$ , between 19.1k $\Omega$  and 232k $\Omega$ . However, if the ILIM pin is connected to VIN, the current limit threshold will be 75mA (typ.). The maximum -100mA fault flag assertion offset needs cautions, especially for very low ILIM applications. Taking the application of ILIM = 250mA as an example, the minimum fault flag assertion level might be 150mA (40% error versus its target). For short ILIM to VIN condition (75mA) the fault flag may go low. Once the current limit threshold is exceeded, the device enters constant current mode until either thermal shutdown occurs or the fault is removed. The table1 shows a recommended current limit value vs. RILIM resistor.

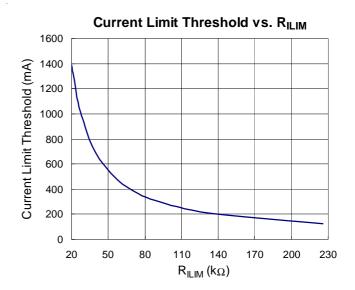


Figure 1. Current Limit Threshold vs. RILIM

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Desired Nominal	Ideal Resistor	Closet 1%	Actual Limits (Include R Tolerance)			
Current Limit (mA)	(kΩ)	Resistor (k $\Omega$ )	IOS Min (mA)	IOS Nom (mA)	IOS Max (mA)	
75	Short Ili	m to VIN	50.0	75.0	100.0	
120	226.1	226.0	101.3	120.0	142.1	
200	134.0	133.0	173.7	201.5	233.9	
300	88.5	88.7	262.1	299.4	342.3	
400	65.9	66.5	351.1	396.7	448.7	
500	52.5	52.3	443.9	501.6	562.4	
600	43.5	43.2	535.1	604.6	674.1	
700	37.2	37.4	616.0	696.0	776.0	
800	32.4	32.4	708.7	800.8	892.9	
900	28.7	28.7	797.8	901.5	1005.2	
1000	25.8	26.1	875.4	989.1	1102.8	
1100	23.4	23.2	982.1	1109.7	1237.3	
1200	21.4	21.5	1057.9	1195.4	1332.9	
1300	19.7	19.6	1158.0	1308.5	1459.0	

#### Table 1. Recommended RILIM Resistor Selections

### Fault Flag

The RT9728 provides a FAULT signal pin which is an N-Channel open drain MOSFET output. This open drain output goes low when current exceeds current limit threshold. The FAULT output is capable of sinking a 1mA load to typically 180mV above ground. The FAULT pin requires a pull-up resistor; this resistor should be large in value to reduce energy drain. A 100k $\Omega$  pull-up resistor works well for most applications. In case of an over-current condition, FAULT will be asserted only after the flag response delay time, t<sub>D</sub>, has elapsed. This ensures that FAULT is asserted upon valid over-current conditions and that erroneous error reporting is eliminated. For example, false over-current conditions may occur during hot-plug events when extremely large capacitive loads are connected, which induces a high transient inrush current that exceeds the current limit threshold. The FAULT response delay time, t<sub>D</sub>, is typically 7.5ms.

#### Supply Filter/Bypass Capacitor

A 10 $\mu$ F low-ESR ceramic capacitor connected from V<sub>IN</sub> to GND and located close to the device is strongly recommended to prevent input voltage drooping during hotplug events. However, higher capacitor values may be used to further reduce the voltage droop on the input. Without this bypass capacitor, an output short may cause sufficient ringing on the input (from source lead inductance) to destroy the internal control circuitry. Note that the input transient voltage must never exceed 6V as stated in the Absolute Maximum Ratings.

### **Output Filter Capacitor**

Standard bypass methods should be used to minimize inductance and resistance between the bypass capacitor and the downstream connector to reduce EMI and decouple voltage droop caused by hot-insertion transients in downstream cables. Ferrite beads in series with V<sub>BUS</sub>, the ground line and the bypass capacitors at the power connector pins are recommended for EMI and ESD protection. The bypass capacitor itself should have a low dissipation factor to allow decoupling at higher frequencies.

For commercial applications where the ambient temperature is  $0^{\circ}$ C to  $70^{\circ}$ C (such as a PC or USB hub),

RT9728 supports an output capacitor range of up to  $120\mu$ F. For industrial applications with an ambient temperature of  $-40^{\circ}$ C to  $125^{\circ}$ C, please limit the output capacitance to less than  $50\mu$ F to ensure normal startup.

### Chip Enable Input

The RT9728 will be disabled when the EN pin is in a logic low condition. During this condition, the internal circuitry and MOSFET are turned off, reducing the supply current to 1 $\mu$ A typical. The maximum guaranteed voltage for a logic-low at the EN pin is 0.66V. A minimum guaranteed voltage of 1.1V at the EN pin will turn on the RT9728. Floating the input may cause unpredictable operation and the EN should not be allowed to go negative with respect to GND. The EN signal must be asserted after input voltage ready or higher than UVLO threshold to satisfy the power sequence.

#### **Under-Voltage Lockout**

Under-Voltage Lockout (UVLO) prevents the MOSFET switch from turning on until input voltage exceeds approximately 2.35V (typ.). If input voltage drops below approximately 2.1V (typ.), UVLO turns off the MOSFET switch.

### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

#### $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = \left(\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}\right) / \theta_{\mathsf{J}\mathsf{A}}$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For WDFN-6L 2x2 packages, the thermal resistance,  $\theta_{JA}$ , is 165°C/W on a standard JEDEC 51-3 single layer thermal test board. The maximum power dissipation at  $T_A = 25^{\circ}C$ 

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can be calculated by the following formula :

 $P_{D(MAX)}$  = (125°C - 25°C) / (165°C/W) = 0.606W for WDFN-6L 2x2 package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . The derating curve in Figure 2 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

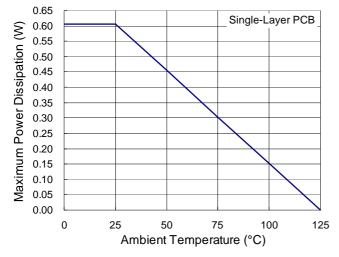
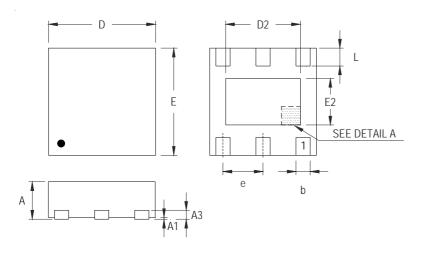
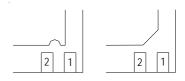


Figure 2. Derating Curve of Maximum Power Dissipation

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## **Outline Dimension**





DETAIL A Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions	In Millimeters	Dimensions In Inches		
	Min	Max	Min	Max	
A	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.200	0.350	0.008	0.014	
D	1.950	2.050	0.077	0.081	
D2	1.000	1.450	0.039	0.057	
E	1.950	2.050	0.077	0.081	
E2	0.500	0.850	0.020	0.033	
е	0.6	650	0.0	)26	
L	0.300	0.400	0.012	0.016	

W-Type 6L DFN 2x2 Package

### **Richtek Technology Corporation**

14F, No. 8, Tai Yuen 1<sup>st</sup> Street, Chupei City Hsinchu, Taiwan, R.O.C. Tel: (8863)5526789

#### RICHTEK

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### **Datasheet Revision History**

Version	Date	Description	Item
10	2023/2/1	Modify	Features on P1 Functional Pin Description on P2

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