General Description

The RT6206A is a synchronous step-down DC/DC converter with Advanced Constant On-Time (ACOT™) mode control. It achieves high power density to deliver up to 5.5A output current from a 4.5V to 18V input supply. The proprietary ACOT™ mode offers an optimal transient response over a wide range of loads and all kinds of ceramic capacitors, which allows the device to adopt very low ESR output capacitor for ensuring performance stabilization. In addition, RT6206A keeps an excellent constant switching frequency under line and load variation and the integrated synchronous power switches with the ACOT™ mode operation provides high efficiency in whole output current load range. Cycle-by-cycle current limit provides an accurate protection by a valley detection of low side MOSFET and external soft-start setting eliminates input current surge during startup. Protection functions include thermal shutdown for RT6206A.

The RT6206A are available in the TSSOP-14 (Exposed pad), SOP-8 (Exposed Pad) and WDFN-10L 3x3 packages.

Features

- ACOT™ Mode Enables Fast Transient Response
- 4.5V to 18V Input Voltage Range
- 5.5A Output Current
- 35mΩ Internal Low Side N-MOSFET
- Advanced Constant On-Time Control
- Support All Ceramic Capacitors
- Up to 95% Efficiency
- Adjustable Output Voltage from 0.765V to 7V
- Adjustable Soft-Start
- Cycle-by-Cycle Current Limit
- Input Under-Voltage Lockout
- Thermal Shutdown
- RoHS Compliant and Halogen Free

Applications

- Industrial and Commercial Low Power Systems
- Computer Peripherals
- LCD Monitors and TVs
- Green Electronics/Appliances
- Point of Load Regulation for High-Performance DSPs, FPGAs, and ASICs

Simplified Application Circuit

* : VINR, GND pin for TSSOP-14 (Exposed Pad) only.

* : PGOOD pin for TSSOP-14 (Exposed Pad) and WDFN-10L 3x3 only.
RT6206A

**Ordering Information**

- **Package Type**
  - CP : TSSOP-14 (Exposed Pad)
- **Lead Plating System**
  - G : Green (Halogen Free and Pb Free)
- **UVP Trim Operation**
  - L : Latch-Off
- **PWM/PSM Mode**
  - A : Force-PWM

- **Package Type**
  - SP : SOP-8 (Exposed Pad-Option 2)
  - QW : WDFN-10L 3x3 (W-Type)
- **Lead Plating System**
  - G : Green (Halogen Free and Pb Free)
- **UVP Trim Operation**
  - H : Hiccup
- **PWM/PSM Mode**
  - A : Force-PWM

**Marking Information**

- **RT6206AHGSP**
  - RT6206AHGSP : Product Number
  - RT6206AHGSP : Product Number
  - YMDNN : Date Code

- **RT6206AHGQW**
  - 5K= : Product Code
  - 5K= : Product Code
  - DNN : Date Code

- **RT6206ALGCP**
  - RT6206ALGCP : Product Number
  - RT6206ALGCP : Product Number
  - YMDNN : Date Code

**Pin Configurations**

**TOP VIEW**

- **SOP-8 (Exposed Pad)**
  - EN
  - FB
  - VREG5
  - SS
  - PGND
  - SW
  - VIN
  - BOOT
  - SW

- **WDFN-10L 3x3**
  - EN
  - FB
  - VREG5
  - SS
  - PGOOD
  - GND
  - PGND
  - VIN
  - BOOT
  - SW
  - SW

- **TSSOP-14 (Exposed Pad)**
  - EN
  - FB
  - VREG5
  - SS
  - GND
  - PGOOD
  - EN
  - VINR
  - VIN
  - BOO
  - SW
  - SW

**Note:**

Richtek products are:
- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.
## Functional Pin Description

<table>
<thead>
<tr>
<th>Pin No.</th>
<th>Pin Name</th>
<th>Pin Function</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>TSSOP-14 (Exposed Pad)</strong></td>
<td><strong>SOP-8 (Exposed Pad)</strong></td>
<td><strong>WDFN-10L 3x3</strong></td>
</tr>
<tr>
<td>1</td>
<td>VOUT</td>
<td>Optional Output Voltage Discharge Connection. This open drain output connects to ground when the device is disabled. If output voltage discharge is desired, connect VOUT to the output voltage.</td>
</tr>
<tr>
<td>2</td>
<td>FB</td>
<td>Feedback Voltage Input. It is used to regulate the output of the converter to a set value via an external resistive voltage divider. The feedback threshold voltage is 0.765V typically.</td>
</tr>
<tr>
<td>3</td>
<td>VREG5</td>
<td>Internal Regulator Output. Connect a 1μF capacitor to GND to stabilize output voltage.</td>
</tr>
<tr>
<td>4</td>
<td>SS</td>
<td>Soft-Start Time Setting. Connect an external capacitor between this pin and GND to set the soft-start time.</td>
</tr>
<tr>
<td>5</td>
<td>GND</td>
<td>Analog Ground.</td>
</tr>
<tr>
<td>6</td>
<td>PGOOD</td>
<td>Open Drain Power Good Indicator Output.</td>
</tr>
<tr>
<td>7</td>
<td>EN</td>
<td>Enable Control Input. A logic-high enables the converter; a logic-low forces the IC into shutdown mode reducing the supply current to less than 10μA.</td>
</tr>
<tr>
<td>8, 9, 15 (Exposed Pad)</td>
<td>5, 9 (Exposed Pad)</td>
<td>11 (Exposed Pad)</td>
</tr>
<tr>
<td>10, 11</td>
<td>SW</td>
<td>Switch Node. Connect this pin to an external L-C filter.</td>
</tr>
<tr>
<td>12</td>
<td>BOOT</td>
<td>Bootstrap Supply for High Side Gate Driver. Connect a 0.1μF capacitor between the BOOT and SW pin.</td>
</tr>
<tr>
<td>13</td>
<td>VIN</td>
<td>Power Input. The input voltage range is from 4.5V to 18V. Must bypass with a suitably large (≥ 10μF x 2) ceramic capacitor.</td>
</tr>
<tr>
<td>14</td>
<td>VINR</td>
<td>Internal Linear Regulator Supply Input. For the TSSOP-14 (Exposed Pad) package, VINR supplies power for the internal linear regulator that powers the IC. Connect VIN to the input voltage and bypass to ground with a 0.1μF ceramic capacitor.</td>
</tr>
</tbody>
</table>
Function Block Diagram

* : VINR, GND pin for TSSOP-14 (Exposed Pad) only.
* : PGOOD pin for TSSOP-14 (Exposed Pad) and WDFN-10L 3x3 only.

Operation

The RT6206A is a synchronous step-down converter with advanced constant on-time control mode. Using the ACOT™ control mode can reduce the output capacitance and provide fast transient response. It can minimize the component size without additional external compensation network.

Internal Regulator

The regulator provides 5.1V power to supply the internal control circuit. Connecting a 1μF ceramic capacitor for decoupling and stability is required.

Soft-Start

In order to prevent the converter output voltage from overshooting during the startup period, the soft-start function is necessary. The soft-start time is adjustable and can be set by an external capacitor.

UVLO Protection

To protect the chip from operating at insufficient supply voltage, the UVLO is needed. When the input voltage of VIN is lower than the UVLO falling threshold voltage, the device will be latch-off.

Thermal Shutdown

When the junction temperature exceeds the OTP threshold value, the IC will shut down the switching operation. Once the junction temperature cools down and is lower than the OTP lower threshold, the converter will automatically resume switching.

Power Good (for TSSOP-14 (Exposed Pad) and WDFN-10L 3x3 only)

After soft-start is finished, the power good function will be activated. When the FB is activated, the PGOOD will become an open-drain output. If the FB is below, the PGOOD pin will be pulled low.
Absolute Maximum Ratings  (Note 1)

- Supply Voltage, VIN, VINR
- Switch Voltage, SW
- < 10ns
- BOOT to SW
- EN, VOUT
- Other Pins
- Power Dissipation, PD @ TA = 25°C

**Recommended Operating Conditions**  (Note 3)

- Supply Voltage, VIN
- Junction Temperature Range
- Ambient Temperature Range

Electrical Characteristics

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Test Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Supply Current</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Shutdown Current</td>
<td>I_{SHDN}</td>
<td>V_{EN} = 0V</td>
<td>--</td>
<td>1</td>
<td>10</td>
<td>μA</td>
</tr>
<tr>
<td>Quiescent Current</td>
<td>I_Q</td>
<td>V_{EN} = 5V, V_FB = 0.8V</td>
<td>--</td>
<td>1</td>
<td>1.3</td>
<td>mA</td>
</tr>
<tr>
<td>Logic Threshold</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>EN Input Voltage</td>
<td>Logic-High</td>
<td></td>
<td>2</td>
<td>--</td>
<td>18</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td>Logic-Low</td>
<td></td>
<td>--</td>
<td>--</td>
<td>0.4</td>
<td>V</td>
</tr>
<tr>
<td>V_FB Voltage</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Feedback Threshold Voltage</td>
<td>V_FB</td>
<td>T_A = 25°C</td>
<td>0.757</td>
<td>0.765</td>
<td>0.773</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>T_A = −40°C to 85°C</td>
<td>0.755</td>
<td>--</td>
<td>0.775</td>
<td>V</td>
</tr>
<tr>
<td>Feedback Input Current</td>
<td>I_FB</td>
<td>V_FB = 0.8V</td>
<td>--</td>
<td>0.01</td>
<td>0.1</td>
<td>μA</td>
</tr>
<tr>
<td>V_REG5 Output</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>V_REG5 Output Voltage</td>
<td>V_REGEX</td>
<td>6V ≤ V_IN ≤ 18V, 0 ≤ I_V_REGEX ≤ 5mA</td>
<td>4.8</td>
<td>5.1</td>
<td>5.4</td>
<td>V</td>
</tr>
<tr>
<td>Line Regulation</td>
<td></td>
<td>6V ≤ V_IN ≤ 18V, I_V_REGEX = 5mA</td>
<td>--</td>
<td>--</td>
<td>20</td>
<td>mV</td>
</tr>
</tbody>
</table>
### Absolute Maximum Ratings

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Test Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Load Regulation</td>
<td>$I_{VREGS}$</td>
<td>$0 &lt; I_{VREGS} &lt; 5mA$</td>
<td>--</td>
<td>--</td>
<td>100</td>
<td>mA</td>
</tr>
<tr>
<td>Output Current</td>
<td>$I_{VREGS}$</td>
<td>$V_{IN} = 6V$, $V_{REGS} = 4V$, $T_{A} = 25^\circ C$</td>
<td>--</td>
<td>70</td>
<td>--</td>
<td>mA</td>
</tr>
<tr>
<td>Switch On Resistance High-Side</td>
<td>$R_{DS(ON)}$</td>
<td>$(V_{BOOT} - V_{SW}) = 5.5V$</td>
<td>--</td>
<td>80</td>
<td>--</td>
<td>mΩ</td>
</tr>
<tr>
<td>Switch On Resistance Low-Side</td>
<td>$R_{DS(ON)}$</td>
<td>--</td>
<td>--</td>
<td>35</td>
<td>--</td>
<td></td>
</tr>
<tr>
<td><strong>Current Limit</strong></td>
<td>$I_{LIM}$</td>
<td></td>
<td>5.8</td>
<td>6.9</td>
<td>8.4</td>
<td>A</td>
</tr>
<tr>
<td><strong>Thermal Shutdown</strong></td>
<td>$T_{SD}$</td>
<td>Shutdown Temperature</td>
<td>--</td>
<td>150</td>
<td>--</td>
<td>°C</td>
</tr>
<tr>
<td>Thermal Shutdown Hysteresis</td>
<td>$\Delta T_{SD}$</td>
<td>--</td>
<td>--</td>
<td>20</td>
<td>--</td>
<td></td>
</tr>
<tr>
<td><strong>On-Time Timer Control</strong></td>
<td>$t_{ON}$</td>
<td>$V_{OUT} = 1.05V$</td>
<td>--</td>
<td>135</td>
<td>--</td>
<td>ns</td>
</tr>
<tr>
<td><strong>Soft-Start</strong></td>
<td>$t_{OFF(MIN)}$</td>
<td>$V_{FB} = 0.7V$</td>
<td>--</td>
<td>260</td>
<td>310</td>
<td>ns</td>
</tr>
<tr>
<td>SS Charge Current</td>
<td>$V_{SS} = 0V$</td>
<td></td>
<td>--</td>
<td>6</td>
<td>--</td>
<td>μA</td>
</tr>
<tr>
<td>SS Discharge Current High-Side</td>
<td>$V_{SS} = 0.5V$ (Latch Mode)</td>
<td>0.1</td>
<td>0.2</td>
<td>--</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>SS Discharge Current Low-Side</td>
<td>$V_{SS} = 0.5V$ (Hiccup Mode)</td>
<td>--</td>
<td>0.5</td>
<td>--</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td><strong>UVLO</strong></td>
<td></td>
<td>Wake Up $V_{REGS}$</td>
<td>3.6</td>
<td>3.85</td>
<td>4.1</td>
<td>V</td>
</tr>
<tr>
<td>Hysteresis</td>
<td></td>
<td></td>
<td>0.16</td>
<td>0.35</td>
<td>0.47</td>
<td></td>
</tr>
<tr>
<td><strong>Output Under Voltage and Over Voltage Protection</strong></td>
<td></td>
<td>OVP Detect</td>
<td>115</td>
<td>120</td>
<td>125</td>
<td>%</td>
</tr>
<tr>
<td>OVP Prop Delay</td>
<td></td>
<td>5</td>
<td>--</td>
<td>--</td>
<td>--</td>
<td>μs</td>
</tr>
<tr>
<td>UVP Trip Threshold</td>
<td></td>
<td>65</td>
<td>70</td>
<td>75</td>
<td>--</td>
<td>%</td>
</tr>
<tr>
<td>UVP Hysteresis</td>
<td></td>
<td>10</td>
<td>--</td>
<td>--</td>
<td>--</td>
<td></td>
</tr>
<tr>
<td>UVP Prop Delay</td>
<td></td>
<td>250</td>
<td>--</td>
<td>--</td>
<td>--</td>
<td>μs</td>
</tr>
<tr>
<td>UVP Enable Delay</td>
<td>$t_{UVPEN}$</td>
<td>Relative to Soft-Start Time</td>
<td>--</td>
<td>$t_{SS} \times 1.7$</td>
<td>--</td>
<td>ms</td>
</tr>
<tr>
<td><strong>Power Good</strong></td>
<td></td>
<td>$V_{FB}$ Rising</td>
<td>85</td>
<td>90</td>
<td>95</td>
<td>%</td>
</tr>
<tr>
<td>PGOOD Threshold</td>
<td></td>
<td>$V_{FB}$ Falling</td>
<td>--</td>
<td>85</td>
<td>--</td>
<td></td>
</tr>
<tr>
<td>PGOOD Sink Current</td>
<td></td>
<td>PGOOD = 0.5V</td>
<td>2.5</td>
<td>5</td>
<td>--</td>
<td>mA</td>
</tr>
</tbody>
</table>

**Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

**Note 2.** $\theta_{JA}$ is measured at $T_{A} = 25^\circ C$ on a high effective thermal conductivity four-layer test board per JEDEC 51-7. $\theta_{JC}$ is measured at the exposed pad of the package.

**Note 3.** Devices are ESD sensitive. Handling precaution is recommended.

**Note 4.** The device is not guaranteed to function outside its operating conditions.
Typical Application Circuit

* : VINR, GND pin for TSSOP-14 (Exposed Pad) only.

* : PGOOD pin for TSSOP-14 (Exposed Pad) and WDFN-10L 3x3 only.

Table 1. Suggested Component Values (VIN = 12V)

<table>
<thead>
<tr>
<th>VOUT (V)</th>
<th>R1 (kΩ)</th>
<th>R2 (kΩ)</th>
<th>C3 (pF)</th>
<th>L1 (µH)</th>
<th>C7 (µF)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>6.81</td>
<td>22.1</td>
<td>--</td>
<td>1.4</td>
<td>22 to 68</td>
</tr>
<tr>
<td>1.05</td>
<td>8.25</td>
<td>22.1</td>
<td>--</td>
<td>1.4</td>
<td>22 to 68</td>
</tr>
<tr>
<td>1.2</td>
<td>12.7</td>
<td>22.1</td>
<td>--</td>
<td>1.4</td>
<td>22 to 68</td>
</tr>
<tr>
<td>1.8</td>
<td>30.1</td>
<td>22.1</td>
<td>5 to 22</td>
<td>2</td>
<td>22 to 68</td>
</tr>
<tr>
<td>2.5</td>
<td>49.9</td>
<td>22.1</td>
<td>5 to 22</td>
<td>2</td>
<td>22 to 68</td>
</tr>
<tr>
<td>3.3</td>
<td>73.2</td>
<td>22.1</td>
<td>5 to 22</td>
<td>2</td>
<td>22 to 68</td>
</tr>
<tr>
<td>5</td>
<td>124</td>
<td>22.1</td>
<td>5 to 22</td>
<td>3.3</td>
<td>22 to 68</td>
</tr>
<tr>
<td>7</td>
<td>180</td>
<td>22.1</td>
<td>5 to 22</td>
<td>3.3</td>
<td>22 to 68</td>
</tr>
</tbody>
</table>
Typical Operating Characteristics

- **Efficiency vs. Output Current**
  - Efficiency (%) vs. Output Current (A)
  - **V_{IN} = 12V**
  - **V_{OUT} = 5V**
  - **V_{OUT} = 1.05V**

- **Output Voltage vs. Output Current**
  - Output Voltage (V) vs. Output Current (A)
  - **V_{IN} = 12V**
  - **V_{OUT} = 1.05V**
  - **I_{OUT} = 0A to 5.5A**

- **Switching Frequency vs. Output Current**
  - Switching Frequency (kHz) vs. Output Current (A)
  - **V_{IN} = 12V**
  - **V_{OUT} = 1.05V**
  - **I_{OUT} = 0A to 5.5A**

- **Feedback Voltage vs. Input Voltage**
  - Feedback Voltage (V) vs. Input Voltage (V)
  - **V_{IN} = 12V**
  - **V_{OUT} = 0.765V**
  - **I_{OUT} = 0.6A**

- **Switching Frequency vs. Temperature**
  - Switching Frequency (kHz) vs. Temperature (°C)

- **Feedback Voltage vs. Temperature**
  - Feedback Voltage (V) vs. Temperature (°C)
  - **V_{IN} = 12V**
  - **V_{OUT} = 0.765V**
  - **I_{OUT} = 0.6A**
### Current Limit vs. Input Voltage

- **Input Voltage (V):** 4.0 to 18.0
- **Current Limit (A):** 5.0 to 9.0

![Current Limit vs. Input Voltage Graph]

### Quiescent Current vs. Temperature

- **Temperature (°C):** -50 to 125
- **Quiescent Current (μA):** 600 to 950

![Quiescent Current vs. Temperature Graph]

### Shutdown Current vs. Temperature

- **Temperature (°C):** -50 to 125
- **Shutdown Current (μA):** 0 to 30

![Shutdown Current vs. Temperature Graph]

### Output Ripple Voltage

- **VIN = 12V, VOUT = 1.05V, IOUT = 5A**
- **VOUT:** 10mV/Div
- **VSW:** 5V/Div

![Output Ripple Voltage Graph]

### Load Transient Response

- **VIN = 12V, VOUT = 1.05V, IOUT = 1A to 5A**
- **VOUT:** 20mV/Div
- **IOUT:** 2A/Div

![Load Transient Response Graph]

### Power On from VIN

- **VIN = 12V, VOUT = 1.05V, IOUT = 5A**
- **VIN:** 10V/Div
- **VOUT:** 0.5V/Div

![Power On from VIN Graph]
UVLO vs. Temperature

EN Threshold Voltage vs. Temperature

Power Off from VIN

Power Off from EN

Power On from EN
Application Information

The RT6206A is a synchronous high voltage Buck converter that can support the input voltage range from 4.5V to 18V and the output current up to 5.5A. It adopts ACOT™ mode control to provide a very fast transient response with few external compensation components.

PWM Operation

It is suitable for low external component count configuration with appropriate amount of Equivalent Series Resistance (ESR) capacitors at the output. The output ripple valley voltage is monitored at a feedback point voltage. The synchronous high side MOSFET is turned on at the beginning of each cycle. After the internal on-time expires, the MOSFET is turned off. The pulse width of this on-time is determined by the converter's input and output voltages to keep the frequency fairly constant over the entire input voltage range.

Advanced Constant On-Time Control

The RT6206A has a unique circuit which sets the on-time by monitoring the input voltage and SW signal. The circuit ensures the switching frequency operating at 650kHz over the entire input voltage range.

Soft-Start

The RT6206A contains an external soft-start clamp that gradually raises the output voltage. The soft-start timing is the output voltage rising time from 0V to settled level and can be programmed by the external capacitor between the SS and GND pins. The chip provides a 6μA charge current for the external capacitor. If a 3.9nF capacitor is used, the soft-start will be 0.5ms (typ.). The available capacitance range is from 2.7nF to 220nF.

\[ t_{SS} (ms) = \frac{C_5 \times 0.765V}{I_{SS} (\mu A)} \]

Chip Enable Operation

The EN pin is the chip enable input. Pulling the EN pin low (<0.4V) will shut down the device. During shutdown mode, the RT6206A’s quiescent current drops to lower than 10μA. Driving the EN pin high (>2V, <18V) will turn on the device again. For external timing control, the EN pin can also be externally pulled high by adding a R_{EN} resistor and C_{EN} capacitor from the VIN pin (see Figure 1).

An external MOSFET can be added to implement digital control on the EN pin when no system voltage above 2V is available, as shown in Figure 2. In this case, a 100kΩ pull-up resistor, R_{EN}, is connected between the VIN and EN pins. MOSFET Q1 will be under logic control to pull down the EN pin.

To prevent enabling circuit when VIN is smaller than the V_OUT target value, a resistive voltage divider can be placed between the input voltage and ground and connected to the EN pin to adjust IC lockout threshold, as shown in Figure 3. For example, if an 8V output voltage is regulated from a 12V input voltage, the resistor R_{EN2} can be selected to set input lockout threshold larger than 8V.
Output Voltage Setting
The resistive divider allows the FB pin to sense the output voltage as shown in Figure 4.

![Figure 4. Output Voltage Setting](image)

The output voltage is set by an external resistive divider according to the following equation. It is recommended to use 1% tolerance or better divider resistors.

\[
V_{\text{OUT}} = 0.765 \times \left(1 + \frac{R_1}{R_2}\right)
\]

Under Voltage Lockout Protection
The RT6206A has Under Voltage Lockout Protection (UVLO) that monitors the voltage of VIN pin. When the VIN voltage is lower than UVLO threshold voltage, the RT6206A will be turned off in this state. This is non-latch protection.

Over Temperature Protection
The RT6206A equips an Over Temperature Protection (OTP) circuitry to prevent overheating due to excessive power dissipation. The OTP will shut down switching operation when junction temperature exceeds 150°C. Once the junction temperature cools down by approximately 25°C the main converter will resume operation. To keep operating at maximum, the junction temperature should be prevented from rising above 150°C.

Hiccup Mode UVP
A Hiccup Mode Under-Voltage Protection (UVP) function is provided for the SOP-8 (Exposed Pad) and WDFN-10L 3x3 packages. When the FB voltage drops below half of the feedback reference voltage, V_FB, the UVP function will be triggered and the RT6206A will shut down for a period of time before recovering automatically. The Hiccup Mode UVP can reduce input current in short-circuit conditions.

Latch off Mode UVP
The Latch off Under-Voltage Protection(UVP) function is provided for the TSSOP-14 (Exposed Pad) package. When the protection function is triggered, the IC will shutdown in Latch-Off Mode. The IC stops switching, leaving both switches open, and is latched off. To restart operation, toggle EN or power the IC off and then on again.

Inductor Selection
The inductor value and operating frequency determine the ripple current according to a specific input and an output voltage. The ripple current \(\Delta I_L\) increases with higher \(V_{\text{IN}}\) and decreases with higher inductance.

\[
\Delta I_L = \frac{V_{\text{OUT}}}{f \times L_{\text{(MAX)}}} \times \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}}(\text{MAX})}\right)
\]

Having a lower ripple current reduces not only the ESR losses in the output capacitors but also the output voltage ripple. High frequency with small ripple current can achieve highest efficiency operation. However, it requires a large inductor to achieve this goal. For the ripple current selection, the value of \(\Delta I_L = 0.2(I_{\text{MAX}})\) will be a reasonable starting point. The largest ripple current occurs at the highest \(V_{\text{IN}}\). To guarantee that the ripple current stays below the specified maximum, the inductor value should be chosen according to the following equation :

\[
L = \frac{V_{\text{OUT}}}{f \times I_{\text{L(MAX)}}} \times \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}(\text{MAX})}}\right)
\]

Input and Output Capacitors Selection
The input capacitance, \(C_{\text{IN}}\), is needed to filter the trapezoidal current at the source of the high side MOSFET. A low ESR input capacitor with larger ripple current rating should be used for the maximum RMS current. The RMS current is given by :

\[
I_{\text{RMS}} = \frac{I_{\text{OUT(MAX)}} \times V_{\text{OUT}}}{V_{\text{IN}}} \sqrt{\frac{V_{\text{IN}}}{V_{\text{OUT}}}}^{-1}
\]

This formula has a maximum at \(V_{\text{IN}} = 2V_{\text{OUT}}\), where \(I_{\text{RMS}} = I_{\text{OUT}} / 2\). This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief.

Choose a capacitor rated at a higher temperature than required. Several capacitors may also be paralleled to meet size or height requirements in the design. For the
input capacitor, two 10μF and 0.1μF low ESR ceramic capacitors are recommended.

The selection of COUT is determined by the required ESR to minimize voltage ripple.

Moreover, the amount of bulk capacitance is also a key for COUT selection to ensure that the control loop is stable. The output ripple, $\Delta V_{OUT}$, is determined by:

$$\Delta V_{OUT} \leq \Delta I_L \left[ \frac{ESR + \frac{1}{2fC_{OUT}}}{E}ight]$$

The output ripple will be highest at the maximum input voltage since $\Delta I_L$ increases with input voltage. Multiple capacitors placed in parallel may need to meet the ESR and RMS current handling requirements.

Higher values, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. However, care must be taken when these capacitors are used at input and output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input, $V_N$. A sudden inrush of current through the long wires can potentially cause a voltage spike at $V_{IN}$ large enough to damage the part.

**External Bootstrap Diode**

Connect a 0.1μF low ESR ceramic capacitor between the BOOT and SW pins. This capacitor provides the gate driver voltage for the high side MOSFET. It is recommended to add an external bootstrap diode between an external 5V and the BOOT pin for efficiency improvement when input voltage is lower than 5.5V or duty ratio is higher than 65%.

The bootstrap diode can be a low cost one such as 1N4148 or BAT54. The external 5V can be a 5V fixed input from system or a 5V output of the RT6206A. Note that the external boot voltage must be lower than 5.5V.

**Figure 5. External Bootstrap Diode**

---

**Over Current Protection**

When the output shorts to ground, the inductor current decays very slowly during a single switching cycle. An over current detector is used to monitor inductor current to prevent current runaway. The over current detector monitors the voltage between SW and GND during the low side MOS turn-on state. This is cycle-by-cycle protection.

**Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = \frac{(T_{J(MAX)} - T_A)}{\theta_{JA}}$$

where $T_{J(MAX)}$ is the maximum junction temperature, $T_A$ is the ambient temperature, and $\theta_{JA}$ is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance, $\theta_{JA}$, is layout dependent. For TSSOP-14 (Exposed Pad) package, the thermal resistance, $\theta_{JA}$, is 40°C/W on a standard JEDEC 51-7 four-layer thermal test board. For SOP-8 (Exposed Pad) package, the thermal resistance, $\theta_{JA}$, is 49°C/W on a standard JEDEC 51-7 four-layer thermal test board. For WDFN-10L 3x3 package, the thermal resistance, $\theta_{JA}$, is 60°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at $T_A = 25°C$ can be calculated by the following formulas:

- For TSSOP-14 (Exposed Pad) package:
  $$P_{D(MAX)} = \frac{(125°C - 25°C)}{(40°C/W)} = 2.5W$$

- For SOP-8 (Exposed Pad) package:
  $$P_{D(MAX)} = \frac{(125°C - 25°C)}{(49°C/W)} = 2.041W$$

- For WDFN-10L 3x3 package:
  $$P_{D(MAX)} = \frac{(125°C - 25°C)}{(60°C/W)} = 1.667W$$

The maximum power dissipation depends on operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance, $\theta_{JA}$. The derating curves in Figure 6 allow the
designer to see the effect of rising ambient temperature on the maximum power dissipation.

![Layout Consideration Diagram](image)

**Layout Consideration**

Follow the PCB layout guidelines for optimal performance of the RT6206A

- Keep the traces of the main current paths as short and wide as possible.
- Put the input capacitor as close as possible to the device pins (VIN and GND).
- SW node is with high frequency voltage swing and should be kept at small area. Keep sensitive components away from the SW node to prevent stray capacitive noise pickup.
- Connect feedback network behind the output capacitors. Keep the loop area small. Place the feedback components near the RT6206A FB pin.
- The GND and Exposed Pad should be connected to a strong ground plane for heat sinking and noise protection.

![Derating Curve of Maximum Power Dissipation](image)
Figure 9. PCB Layout Guide for TSSOP-14 (Exposed Pad) Package

Place the feedback components as close to the FB as possible for better regulation.

Place the input and output capacitors as close to the IC as possible.

SW should be connected to inductor by Wide and short trace. Keep sensitive components away from this trace.

Place the input and output capacitors as close to the IC as possible.
Outline Dimension

8-Lead SOP (Exposed Pad) Plastic Package

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Dimensions In Millimeters</th>
<th>Dimensions In Inches</th>
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<tbody>
<tr>
<td></td>
<td>Min</td>
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</tr>
<tr>
<td>B</td>
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<td></td>
<td>Y</td>
<td>3.000</td>
</tr>
</tbody>
</table>
Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.
Symbol | Dimensions In Millimeters | Dimensions In Inches
--- | --- | ---
A | 1.000 to 1.200 | 0.039 to 0.047 |
A1 | 0.000 to 0.150 | 0.000 to 0.006 |
A2 | 0.800 to 1.050 | 0.031 to 0.041 |
b | 0.190 to 0.300 | 0.007 to 0.012 |
D | 4.900 to 5.100 | 0.193 to 0.201 |
e | 0.650 | 0.026 |
E | 6.300 to 6.500 | 0.248 to 0.256 |
E1 | 4.300 to 4.500 | 0.169 to 0.177 |
L | 0.450 to 0.750 | 0.018 to 0.030 |
U | 1.900 to 2.900 | 0.075 to 0.114 |
V | 1.600 to 2.600 | 0.063 to 0.102 |

14-Lead TSSOP (Exposed Pad) Plastic Package

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